

Fundamentals of **LTE**

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Foreword by Rajiv Laroia

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Library of Congress Cataloging-in-Publication Data

Fundamentals of LTE / Arunabha Ghosh . . . [et al.].

p. cm.

Includes bibliographical references and index.

ISBN-10: 0-13-703311-7 (hardcover : alk. paper)

ISBN-13: 978-0-13-703311-9 (hardcover : alk. paper) 1. Long-Term Evolution (Telecommunications) I. Ghosh, Arunabha, 1969–

TK5103.48325.F86 2010

621.3845'6—dc22

2010021369

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Pearson Education, Inc
Rights and Contracts Department
501 Boylston Street, Suite 900
Boston, MA 02116
Fax: (617) 671-3447

ISBN-13: 978-0-13-703311-9

ISBN-10: 0-13-703311-7

Text printed in the United States on recycled paper at Courier in Westford, Massachusetts.
First printing, August 2010

Contents

Foreword	xvii
Preface	xix
Acknowledgments	xxi
About the Authors	xxiii
List of Acronyms	xxv
1 Evolution of Cellular Technologies	1
1.1 Introduction	1
1.2 Evolution of Mobile Broadband	3
1.2.1 First Generation Cellular Systems	5
1.2.2 2G Digital Cellular Systems	6
1.2.3 3G Broadband Wireless Systems	10
1.2.4 Beyond 3G: HSPA+, WiMAX, and LTE	15
1.2.5 Summary of Evolution of 3GPP Standards	22
1.3 The Case for LTE/SAE	23
1.3.1 Demand Drivers for LTE	24
1.3.2 Key Requirements of LTE Design	26
1.4 Key Enabling Technologies and Features of LTE	28
1.4.1 Orthogonal Frequency Division Multiplexing (OFDM)	28
1.4.2 SC-FDE and SC-FDMA	30
1.4.3 Channel Dependent Multi-user Resource Scheduling	30
1.4.4 Multiantenna Techniques	31
1.4.5 IP-Based Flat Network Architecture	32

1.5	LTE Network Architecture	33
1.6	Spectrum Options and Migration Plans for LTE	35
1.7	Future of Mobile Broadband—Beyond LTE	39
1.8	Summary and Conclusions	41
Part I LTE Tutorials		45
2	Wireless Fundamentals	47
2.1	Communication System Building Blocks	47
2.2	The Broadband Wireless Channel: Path Loss and Shadowing	48
2.2.1	Path Loss	50
2.2.2	Shadowing	53
2.3	Cellular Systems	56
2.3.1	The Cellular Concept	57
2.3.2	Analysis of Cellular Systems	58
2.3.3	Sectoring	60
2.4	The Broadband Wireless Channel: Fading	62
2.4.1	Delay Spread and Coherence Bandwidth	66
2.4.2	Doppler Spread and Coherence Time	67
2.4.3	Angular Spread and Coherence Distance	68
2.5	Modelling Broadband Fading Channels	69
2.5.1	Statistical Channel Models	70
2.5.2	Statistical Correlation of the Received Signal	73
2.5.3	Empirical Channel Models	77
2.6	Mitigation of Narrowband Fading	82
2.6.1	The Effects of Unmitigated Fading	82
2.6.2	Spatial Diversity	84
2.6.3	Coding and Interleaving	85
2.6.4	Automatic Repeat Request (ARQ)	88
2.6.5	Adaptive Modulation and Coding (AMC)	88
2.6.6	Combining Narrowband Diversity Techniques—The Whole Is Less Than the Sum of the Parts	91
2.7	Mitigation of Broadband Fading	92
2.7.1	Spread Spectrum and RAKE Receivers	92
2.7.2	Equalization	93
2.7.3	Multicarrier Modulation: OFDM	93
2.7.4	Single-Carrier Modulation with Frequency Domain Equalization	94
2.8	Chapter Summary	94

3	Multicarrier Modulation	99
3.1	The Multicarrier Concept	100
3.1.1	An Elegant Approach to Intersymbol Interference	101
3.2	OFDM Basics	103
3.2.1	Block Transmission with Guard Intervals	103
3.2.2	Circular Convolution and the DFT	104
3.2.3	The Cyclic Prefix	105
3.2.4	Frequency Equalization	107
3.2.5	An OFDM Block Diagram	108
3.3	OFDM in LTE	109
3.4	Timing and Frequency Synchronization	110
3.4.1	Timing Synchronization	111
3.4.2	Frequency Synchronization	114
3.5	The Peak-to-Average Ratio	116
3.5.1	The PAR Problem	116
3.5.2	Quantifying the PAR	118
3.5.3	Clipping and Other PAR Reduction Techniques	121
3.5.4	LTE's Approach to PAR in the Uplink	123
3.6	Single-Carrier Frequency Domain Equalization (SC-FDE)	124
3.6.1	SC-FDE System Description	124
3.6.2	SC-FDE Performance vs. OFDM	126
3.6.3	Design Considerations for SC-FDE and OFDM	126
3.7	The Computational Complexity Advantage of OFDM and SC-FDE	127
3.8	Chapter Summary	130
4	Frequency Domain Multiple Access: OFDMA and SC-FDMA	133
4.1	Multiple Access for OFDM Systems	134
4.1.1	Multiple Access Overview	134
4.1.2	Random Access vs. Multiple Access	135
4.1.3	Frequency Division Multiple Access (OFDM-FDMA)	136
4.1.4	Time Division Multiple Access (OFDM-TDMA)	137
4.1.5	Code Division Multiple Access (OFDM-CDMA or MC-CDMA)	137
4.2	Orthogonal Frequency Division Multiple Access (OFDMA)	138
4.2.1	OFDMA: How It Works	139
4.2.2	OFDMA Advantages and Disadvantages	142
4.3	Single-Carrier Frequency Division Multiple Access (SC-FDMA)	142

4.3.1	SC-FDMA: How It Works	142
4.3.2	SC-FDMA Advantages and Disadvantages	143
4.4	Multiuser Diversity and Opportunistic Scheduling	144
4.4.1	Multiuser Diversity	144
4.4.2	Opportunistic Scheduling Approaches for OFDMA	146
4.4.3	Maximum Sum Rate Algorithm	147
4.4.4	Maximum Fairness Algorithm	148
4.4.5	Proportional Rate Constraints Algorithm	148
4.4.6	Proportional Fairness Scheduling	149
4.4.7	Performance Comparison	150
4.5	OFDMA and SC-FDMA in LTE	152
4.5.1	The LTE Time-Frequency Grid	153
4.5.2	Allocation Notification and Uplink Feedback	154
4.5.3	Power Control	154
4.6	OFDMA System Design Considerations	155
4.6.1	Resource Allocation in Cellular Systems	156
4.6.2	Fractional Frequency Reuse in Cellular Systems	157
4.6.3	Multiuser Diversity vs. Frequency and Spatial Diversity	159
4.7	Chapter Summary	160
5	Multiple Antenna Transmission and Reception	167
5.1	Spatial Diversity Overview	168
5.1.1	Array Gain	168
5.1.2	Diversity Gain	169
5.1.3	Increasing the Data Rate with Spatial Diversity	170
5.1.4	Increased Coverage or Reduced Transmit Power	171
5.2	Receive Diversity	171
5.2.1	Selection Combining	171
5.2.2	Maximal Ratio Combining	173
5.3	Transmit Diversity	174
5.3.1	Open-Loop Transmit Diversity: 2×1 Space-Frequency Block Coding	175
5.3.2	Open-Loop Transmit Diversity with More Antennas	177
5.3.3	Transmit Diversity vs. Receive Diversity	180
5.3.4	Closed-Loop Transmit Diversity	181
5.4	Interference Cancellation Suppression and Signal Enhancement	186

5.4.1	DOA-Based Beamsteering	187
5.4.2	Linear Interference Suppression: Complete Knowledge of Interference Channels	189
5.4.3	Linear Interference Suppression: Statistical Knowledge of Interference Channels	190
5.5	Spatial Multiplexing	192
5.5.1	An Introduction to Spatial Multiplexing	193
5.5.2	Open-Loop MIMO: Spatial Multiplexing Without Channel Feedback	194
5.5.3	Closed-Loop MIMO	198
5.6	How to Choose Between Diversity, Interference Suppression, and Spatial Multiplexing	200
5.7	Channel Estimation and Feedback for MIMO and MIMO-OFDM	202
5.7.1	Channel Estimation	202
5.7.2	Channel Feedback	206
5.8	Practical Issues That Limit MIMO Gains	208
5.8.1	Multipath	208
5.8.2	Uncorrelated Antennas	208
5.8.3	Interference-Limited MIMO Systems	209
5.9	Multiuser and Networked MIMO Systems	209
5.9.1	Multiuser MIMO	210
5.9.2	Networked MIMO	212
5.10	An Overview of MIMO in LTE	213
5.10.1	An Overview of MIMO in the LTE Downlink	213
5.10.2	An Overview of MIMO in the LTE Uplink	214
5.11	Chapter Summary	215
Part II The LTE Standard		225
6	Overview and Channel Structure of LTE	227
6.1	Introduction to LTE	228
6.1.1	Design Principles	229
6.1.2	Network Architecture	231
6.1.3	Radio Interface Protocols	232
6.2	Hierarchical Channel Structure of LTE	234
6.2.1	Logical Channels: What to Transmit	235
6.2.2	Transport Channels: How to Transmit	236

6.2.3	Physical Channels: Actual Transmission	239
6.2.4	Channel Mapping	240
6.3	Downlink OFDMA Radio Resources	241
6.3.1	Frame Structure	242
6.3.2	Physical Resource Blocks for OFDMA	246
6.3.3	Resource Allocation	248
6.3.4	Supported MIMO Modes	251
6.4	Uplink SC-FDMA Radio Resources	251
6.4.1	Frame Structure	252
6.4.2	Physical Resource Blocks for SC-FDMA	252
6.4.3	Resource Allocation	254
6.4.4	Supported MIMO Modes	254
6.5	Summary and Conclusions	255
7	Downlink Transport Channel Processing	257
7.1	Downlink Transport Channel Processing Overview	257
7.1.1	Channel Coding Processing	258
7.1.2	Modulation Processing	263
7.2	Downlink Shared Channels	268
7.2.1	Channel Encoding and Modulation	269
7.2.2	Multiantenna Transmission	270
7.3	Downlink Control Channels	276
7.3.1	Downlink Control Information (DCI) Formats	277
7.3.2	Channel Encoding and Modulation	279
7.3.3	Multiantenna Transmission	282
7.4	Broadcast Channels	283
7.5	Multicast Channels	284
7.6	Downlink Physical Signals	285
7.6.1	Downlink Reference Signals	285
7.6.2	Synchronization Signals	289
7.7	H-ARQ in the Downlink	290
7.8	Summary and Conclusions	293
8	Uplink Transport Channel Processing	295
8.1	Uplink Transport Channel Processing Overview	296
8.1.1	Channel Coding Processing	296
8.1.2	Modulation Processing	297

8.2	Uplink Shared Channels	298
8.2.1	Channel Encoding and Modulation	299
8.2.2	Frequency Hopping	299
8.2.3	Multiantenna Transmission	300
8.3	Uplink Control Information	301
8.3.1	Channel Coding for Uplink Control Information	302
8.3.2	Modulation of PUCCH	306
8.3.3	Resource Mapping	308
8.4	Uplink Reference Signals	309
8.4.1	Reference Signal Sequence	310
8.4.2	Resource Mapping of Demodulation Reference Signals	310
8.4.3	Resource Mapping of Sounding Reference Signals	311
8.5	Random Access Channels	313
8.6	H-ARQ in the Uplink	315
8.6.1	The FDD Mode	315
8.6.2	The TDD Mode	315
8.7	Summary and Conclusions	317
9	Physical Layer Procedures and Scheduling	319
9.1	Hybrid-ARQ Feedback	319
9.1.1	H-ARQ Feedback for Downlink (DL) Transmission	320
9.1.2	H-ARQ Indicator for Uplink (UL) Transmission	321
9.2	Channel Quality Indicator (CQI) Feedback	322
9.2.1	A Primer on CQI Estimation	323
9.2.2	CQI Feedback Modes	325
9.3	Precoder for Closed-Loop MIMO Operations	333
9.3.1	Precoder Estimation for Multicarrier Systems	333
9.3.2	Precoding Matrix Index (PMI) and Rank Indication (RI) Feedback	334
9.4	Uplink Channel Sounding	337
9.5	Buffer Status Reporting in Uplink	337
9.6	Scheduling and Resource Allocation	339
9.6.1	Signaling for Scheduling in Downlink and Uplink	340
9.6.2	Multuser MIMO Signaling	344
9.7	Semi-persistent Scheduling for VoIP	344
9.7.1	Motivation for Semi-persistent Scheduling	344
9.7.2	Changes in the Signaling Structure	345

9.8	Cell Search	346
9.9	Random Access Procedures	348
9.10	Power Control in Uplink	350
9.11	Summary and Conclusions	352
10	Data Flow, Radio Resource Management, and Mobility Management	355
10.1	PDCP Overview	359
10.1.1	Header Compression	361
10.1.2	Integrity and Ciphering	361
10.2	MAC/RLC Overview	363
10.2.1	Data Transfer Modes	363
10.2.2	Purpose of MAC and RLC Layers	364
10.2.3	PDU Headers and Formats	365
10.2.4	ARQ Procedures	368
10.3	RRC Overview	369
10.3.1	RRC States	369
10.3.2	RRC Functions	370
10.4	Mobility Management	371
10.4.1	S1 Mobility	371
10.4.2	X2 Mobility	373
10.4.3	RAN Procedures for Mobility	374
10.4.4	Paging	376
10.5	Inter-cell Interference Coordination	377
10.5.1	Downlink	377
10.5.2	Uplink	379
10.6	Summary and Conclusions	380
	Index	383

Foreword

With the deployment of LTE, the wireless revolution will achieve an important milestone. For the first time, a wide-area wireless network will be universally deployed that has been primarily designed for IP-centric broadband data (rather than voice) from the very beginning. LTE also is rapidly becoming the dominant global standard for fourth generation cellular networks with nearly all the major cellular players behind it and working toward its success.

Having been personally involved in designing, developing, and promoting one of the first OFDM-based cellular systems since the late 1990s, back when such an approach was considered slightly eccentric, LTE's success is personally very satisfying for me to see. As with any standard, which by political necessity is "designed by committee," the LTE specification is not without flaws and there is room for progress and future evolution. The system architecture is not yet a fully flat IP platform, for example, and some interference issues are not fully addressed. But there can be no doubt that LTE is a giant step in the right direction and a necessary step to meet the anticipated growth in consumer and business mobile broadband applications and services. LTE provides a credible platform for wireless broadband access based on OFDMA, multiantenna technologies, and other cutting-edge techniques that provide improvements in spectral efficiency and significantly lower the cost of delivering mobile broadband. I expect the future evolution of LTE to continually improve the standard.

Fundamentals of LTE is an excellent introduction to the LTE standard and the various technologies that it incorporates, like OFDMA, SC-FDMA, and multiantenna transmission and reception. It is exceptionally well written, easy to understand, and concisely but completely covers the key aspects of the standard. Because of its diverse author team—including both LTE systems engineers as well as leading academic researchers who have worked extensively on the core underlying technologies—this book will be of use to a wide set of potential readers. I recommend it to folks in the industry who are involved with the development of LTE-based technology and products, as well as to students and faculty in academia who wish to understand the standard and participate in incorporating more advanced techniques into the future versions of the specification. The book also describes some of the "weak points" in the current specification of the standard. This helps ensure that these issues will be fixed as the specification evolves.

I hope you will enjoy reading the book and benefit from it, and am confident you will.

Rajiv Laroia

Senior vice president, Qualcomm Flarion Technologies

Preface

The Long-Term Evolution (LTE) is the next evolutionary step beyond 3G for mobile wireless communication. LTE brings together many technological innovations from different areas of research such as digital signal processing, Internet protocols, network architecture, and security, and is poised to dramatically change the way we use the worldwide mobile network in the future. Unlike 3G, LTE uses a clean-slate design approach for all the components of the network including the radio access network, the transport network, and the core network. This design approach, along with its built-in flexibility, allows LTE to be the first truly global wireless standard that can be deployed in a variety of spectrum and operating scenarios, and support a wide range of wireless applications. A large number of service providers around the world have already announced LTE as their preferred next generation technology.

Fundamentals of LTE is a comprehensive tutorial on the most innovative cellular standard since CDMA emerged in the early 1990s. The impending worldwide deployment of LTE (Long-Term Evolution, often called 4G cellular) will revolutionize the cellular networks by going to much larger bandwidths, data rates, and an all-IP framework. *Fundamentals of LTE* is the only book to provide an accessible but complete tutorial on the key enabling technologies behind LTE, such as OFDM, OFDMA, SC-FDMA, and MIMO, as well as provide a step-by-step breakdown of all the key aspects of the standard from the physical layer through the network stack. The book begins with a historical overview and the reasons for the radical departure from conventional voice-centric cellular systems that LTE represents. Following this, four tutorial chapters explain the essential underpinnings of LTE, which could also be used as the basis for an entry-level university course. Finally, five chapters on the LTE standard specifically attempt to illuminate its key aspects, explaining both how LTE works, and why certain choices were made by the LTE standards body. This collaboration between UT Austin and AT&T has resulted in a uniquely accessible and comprehensive book on LTE.

Chapter 1 provides an overview and history of the cellular wireless technologies, starting from first-generation systems such as AMPS to fourth-generation technologies such as LTE and WiMAX. This chapter provides a historical account of the mobile wireless networks and illustrates the key technological breakthroughs and market forces that drove the evolution of the mobile wireless network over the past two decades. This chapter also provides an executive summary of the LTE and some of its key technical enablers.

The balance of the book is organized into two parts, as noted. Part I consists of four tutorial chapters (Chapters 2–5) on the essential wireless networking and communications

technologies underpinning LTE. Chapter 2 provides a tutorial introduction to broadband wireless channels and systems, and demonstrates the challenges inherent to the development of a broadband wireless system such as LTE. Chapter 3 provides a comprehensive tutorial on multicarrier modulation, detailing how it works in both theory and practice. This chapter emphasizes a practical understanding of OFDM system design, and discusses implementation issues, in particular the peak-to-average power ratio. An overview of single-carrier frequency domain equalization (SC-FDE), which overcomes the peak-to-average problem, is also provided. Chapter 4 extends Chapter 3 to provide an overview on the frequency domain multiple access techniques adopted in LTE: OFDMA in the downlink and SC-FDMA in the uplink. Resource allocation to the users, especially relevant opportunistic scheduling approaches, is discussed, along with important implementation issues pertinent to LTE. Chapter 5 provides a rigorous tutorial on multiple antenna techniques, covering techniques such as spatial diversity, interference cancellation, spatial multiplexing, and multiuser and networked MIMO. The inherent tradeoffs between different techniques and practical considerations for the deployment of MIMO in LTE are distinguishing features of this chapter.

Part II of the book, consisting of Chapters 6–10, provides a detailed description of the LTE standard with particular emphasis on the air-interface protocol. We begin this part in Chapter 6 with an introduction to the basic structure of the air-interface protocol and the channel structure utilized by LTE at different layers. This chapter also provides an overview of the physical layer and various OFDMA-related aspects of LTE. Chapters 7 and 8 provide a thorough description of the physical and MAC layer processing (at the transport channel level) for downlink (DL) and uplink (UL), respectively. Features such as channel encoding, modulation mapping, Hybrid-ARQ (H-ARQ), and multiantenna processing for the different DL and UL channels are discussed in detail. In Chapter 9 we discuss the various feedback mechanisms that are essential components of LTE and are needed to enable various features such as channel aware scheduling, closed-loop and open-loop multiantenna processing, adaptive modulation and coding, etc. These concepts are critical to a complete understanding of LTE and its operation. In this chapter we also discuss various MAC layer concepts related to scheduling, QoS, ARQ, etc. Finally, in Chapter 10 we discuss the higher layers of the LTE protocol stack, such as RLC, PDCP, and RRM, and the role of these in the overall operation of an LTE system. In this chapter we also provide an in-depth discussion on the mobility and handoff procedures in LTE from a radio access network point of view.

Overview and Channel Structure of LTE

In Part I, we discussed the inherent challenges and associated technical solutions in designing a broadband wireless network. From here onward, we describe the technical details of the LTE specifications. As a starting point, in this chapter we provide an overview of the LTE radio interface. The 3rd Generation Partnership Project (3GPP) defines a separable network structure, that is, it divides the whole network into a radio access network (RAN) and a core network (CN), which makes it feasible to evolve each part independently. The *Long-Term Evolution (LTE)* project in 3GPP focuses on enhancing the UMTS Terrestrial Radio Access (UTRA)—the 3G RAN developed within 3GPP, and on optimizing 3GPP’s overall radio access architecture. Another parallel project in 3GPP is the *Evolved Packet Core (EPC)*, which focuses on the CN evolution with a flatter all-IP, packet-based architecture. The complete packet system consisting of LTE and EPC is called the *Evolved Packet System (EPS)*. This book focuses on LTE, while EPC is discussed only when necessary. LTE is also referred to as *Evolved UMTS Terrestrial Radio Access (E-UTRA)*, and the RAN of LTE is also referred to as *Evolved UMTS Terrestrial Radio Access Network (E-UTRAN)*.

The radio interface of a wireless network is the interface between the mobile terminal and the base station, and thus in the case of LTE it is located between the RAN–E-UTRAN and the user equipment (UE, the name for the mobile terminal in 3GPP). Compared to the UMTS Terrestrial Radio Access Network (UTRAN) for 3G systems, which has two logical entities—the Node-B (the radio base station) and the radio network controller (RNC)—the E-UTRAN network architecture is simpler and flatter. It is composed of only one logical node—the evolved Node-B (eNode-B). The RAN architectures of UTRAN and E-UTRAN are shown in Figure 6.1. Compared to the traditional Node-B, the eNode-B supports additional features, such as radio resource control, admission control, and mobility management, which were originally contained in the RNC. This simpler structure simplifies the network operation and allows for higher throughput and lower latency over the radio interface.

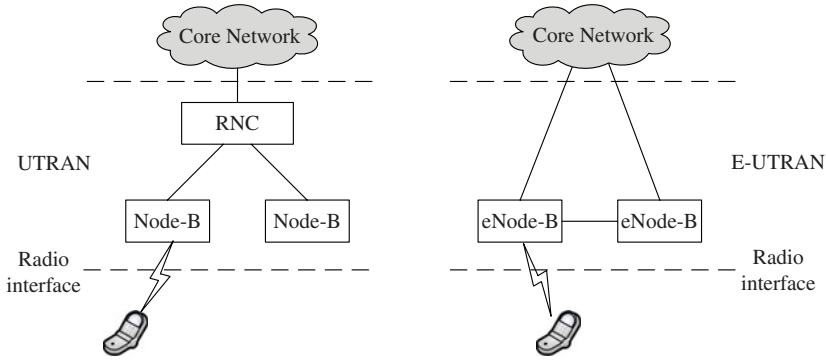


Figure 6.1 Radio interface architectures of UTRAN and E-UTRAN.

The LTE radio interface aims for a *long-term evolution*, so it is designed with a clean slate approach as opposed to High-Speed Packet Access (HSPA), which was designed as an add-on to UMTS in order to increase throughput of packet switched services. HSPA is a collection of *High-Speed Downlink Packet Access (HSDPA)* and *High-Speed Uplink Packet Access (HSUPA)*. The clean slate approach allows for a completely different air interface, which means that advanced techniques, including *Orthogonal Frequency Division Multiplexing (OFDM)* and *multiantenna transmission and reception (MIMO)*, could be included from the start of the standardization of LTE. For multiple access, it moves away from *Code Division Multiple Access (CDMA)* and instead uses *Orthogonal Frequency Division Multiple Access (OFDMA)* in the downlink and *Single-Carrier Frequency Division Multiple Access (SC-FDMA)* in the uplink. All these techniques were described in detail in Part I, so in Part II we assume a basic knowledge of a wireless system, antenna diversity, OFDMA, and other topics covered in Part I.

In this chapter, we provide an introduction to the LTE radio interface, and describe its hierarchical channel structure. First, an overview of the LTE standard is provided, including design principles, the network architecture, and radio interface protocols. We then describe the purpose of each channel type defined in LTE and the mapping between channels at various protocol layers. Next, the downlink OFDMA and uplink SC-FDMA aspects of the air interface are described, including frame structures, physical resource blocks, resource allocation, and the supported MIMO modes. This chapter serves as the foundation for understanding the physical layer procedures and higher layer protocols of LTE that are described in the chapters to follow.

6.1 Introduction to LTE

As mentioned previously, LTE is the next step in the evolution of mobile cellular systems and was standardized as part of the 3GPP Release 8 specifications. Unlike 2G and 3G cellular systems¹ that were designed mainly with voice services in mind, LTE was

¹ Evolution of different 3GPP standards, including GPRS, UMTS, and HSPA, was discussed in Chapter 1.

designed primarily for high-speed data services, which is why LTE is a packet-switched network from end to end and has no support for circuit-switched services. However, the low latency of LTE and its sophisticated quality of service (QoS) architecture allow a network to emulate a circuit-switched connection on top of the packet-switched framework of LTE.

6.1.1 Design Principles

The LTE standard was designed as a completely new standard, with new numbering and new documentation, and it is not built on the previous versions of 3GPP standards. Earlier elements were brought in only if there was a compelling reason for them to exist in the new standard. The basic design principles that were agreed upon and followed in 3GPP while designing the LTE specifications include:²

- **Network Architecture:** Unlike 3G networks, LTE was designed to support packet-switched traffic with support for various QoS classes of services. Previous generations of networks such as UMTS/HSPA and 1xRTT/EvDO also support packet-switched traffic but this was achieved by subsequent add-ons to the initial version of the standards. For example, HSPA, which is a packet-switched protocol (packet-switched over the air), was built on top of the Release 99 UMTS network and as a result carried some of the unnecessary burdens of a circuit-switched network. LTE is different in the sense that it is a clean slate design and supports packet switching for high data rate services from the start. The LTE radio access network, E-UTRAN, was designed to have the minimum number of interfaces (i.e., the minimum number of network elements) while still being able to provide efficient packet-switched transport for traffic belonging to all the QoS classes such as conversational, streaming, real-time, non-real-time, and background classes.
- **Data Rate and Latency:** The design target for downlink and uplink peak data rates for LTE are 100 Mbps and 50 Mbps, respectively, when operating at the 20MHz frequency division duplex (FDD) channel size. The user-plane latency is defined in terms of the time it takes to transmit a small IP packet from the UE to the edge node of the radio access network or vice versa measured on the IP layer. The target for one-way latency in the user plane is 5 ms in an unloaded network, that is, if only a single UE is present in the cell. For the control-plane latency, the transition time from a camped state to an active state is less than 100 ms, while the transition time between a dormant state and an active state should be less than 50 ms.
- **Performance Requirements:** The target performance requirements for LTE are specified in terms of spectrum efficiency, mobility, and coverage, and they are in general expressed relative to the 3GPP Release 6 HSPA.
 - **Spectrum Efficiency** The average downlink user data rate and spectrum efficiency target is three to four times that of the baseline HSDPA network. Similarly, in the uplink the average user data rate and spectrum efficiency

² See Section 1.2.4 for a comparison of different beyond-3G systems, including HSPA+, WiMAX, and LTE.

target is two to three times that of the baseline HSUPA network. The cell edge throughput, measured as the 5th percentile throughput, should be two to three times that of the baseline HSDPA and HSUPA.

- **Mobility** The mobility requirement for LTE is to be able to support hand-off/mobility at different terminal speeds. Maximum performance is expected for the lower terminal speeds of 0 to 15 km/hr, with minor degradation in performance at higher mobile speeds up to 120 km/hr. LTE is also expected to be able to sustain a connection for terminal speeds up to 350 km/hr but with significant degradation in the system performance.
 - **Coverage** For the cell coverage, the above performance targets should be met up to 5 km. For cell ranges up to 30 km, a slight degradation of the user throughput is tolerated and a more significant degradation for spectrum efficiency is acceptable, but the mobility requirements should be met. Cell ranges up to 100 km should not be precluded by the specifications.
 - **MBMS Service** LTE should also provide enhanced support for the Multimedia Broadcast and Multicast Service (MBMS) compared to UTRA operation.
- **Radio Resource Management:** The radio resource management requirements cover various aspects such as enhanced support for end-to-end QoS, efficient support for transmission of higher layers, and support for load sharing/balancing and policy management/enforcement across different radio access technologies.
 - **Deployment Scenario and Co-existence with 3G:** At a high level, LTE shall support the following two deployment scenarios:
 - Standalone deployment scenario, where the operator deploys LTE either with no previous network deployed in the area or with no requirement for interworking with the existing UTRAN/GERAN (GSM EDGE radio access network) networks.
 - Integrating with existing UTRAN and/or GERAN deployment scenario, where the operator already has either a UTRAN and/or a GERAN network deployed with full or partial coverage in the same geographical area.
 - **Flexibility of Spectrum and Deployment:** In order to become a truly global standard, LTE was designed to be operable under a wide variety of spectrum scenarios, including its ability to coexist and share spectrum with existing 3G technologies. Service providers in different geographical regions often have different spectrums in terms of the carrier frequency and total available bandwidth, which is why LTE was designed to have a scalable bandwidth from 1.4MHz to 20MHz. In order to accommodate flexible duplexing options, LTE was designed to operate in both frequency division duplex (FDD) and time division duplex (TDD) modes.
 - **Interoperability with 3G and 2G Networks:** Multimode LTE terminals, which support UTRAN and/or GERAN operation, should be able to support measurement of, and handover from and to, both 3GPP UTRAN and 3GPP GERAN systems with acceptable terminal complexity and network performance.

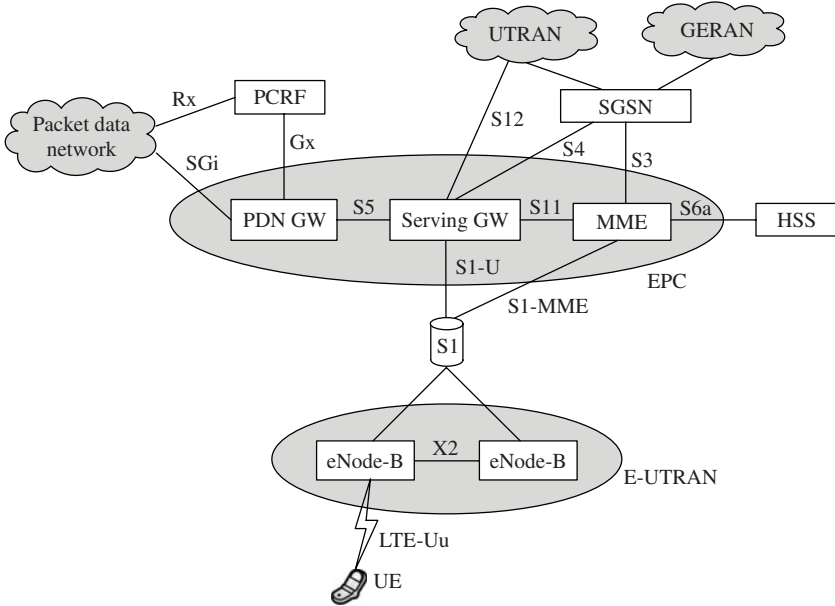


Figure 6.2 LTE end-to-end network architecture.

6.1.2 Network Architecture

Figure 6.2 shows the end-to-end network architecture of LTE and the various components of the network. The entire network is composed of the radio access network (E-UTRAN) and the core network (EPC), both of which have been defined as new components of the end-to-end network in Release 8 of the 3GPP specifications. In this sense, LTE is different from UMTS since UMTS defined a new radio access network but used the same core network as the previous-generation Enhanced GPRS (EDGE) network. This obviously has some implications for the service providers who are upgrading from a UMTS network to LTE. The main components of the E-UTRAN and EPC are

- **UE:** The mobile terminal.
- **eNode-B:** The eNode-B (also called the base station) terminates the air interface protocol and is the first point of contact for the UE. As already shown in Figure 6.1, the eNode-B is the only logical node in the E-UTRAN, so it includes some functions previously defined in the RNC of the UTRAN, such as radio bearer management, uplink and downlink dynamic radio resource management and data packet scheduling, and mobility management.
- **Mobility Management Entity (MME):** MME is similar in function to the control plane of legacy Serving GPRS Support Node (SGSN). It manages mobility aspects in 3GPP access such as gateway selection and tracking area list management.

- **Serving Gateway (Serving GW):** The Serving GW terminates the interface toward E-UTRAN, and routes data packets between E-UTRAN and EPC. In addition, it is the local mobility anchor point for inter-eNode-B handovers and also provides an anchor for inter-3GPP mobility. Other responsibilities include lawful intercept, charging, and some policy enforcement. The Serving GW and the MME may be implemented in one physical node or separate physical nodes.
- **Packet Data Network Gateway (PDN GW):** The PDN GW terminates the SGi interface toward the Packet Data Network (PDN). It routes data packets between the EPC and the external PDN, and is the key node for policy enforcement and charging data collection. It also provides the anchor point for mobility with non-3GPP accesses. The external PDN can be any kind of IP network as well as the IP Multimedia Subsystem (IMS) domain. The PDN GW and the Serving GW may be implemented in one physical node or separated physical nodes.
- **S1 Interface:** The S1 interface is the interface that separates the E-UTRAN and the EPC. It is split into two parts: the S1-U, which carries traffic data between the eNode-B and the Serving GW, and the S1-MME, which is a signaling-only interface between the eNode-B and the MME.
- **X2 Interface:** The X2 interface is the interface between eNode-Bs, consisting of two parts: the X2-C is the control plane interface between eNode-Bs, while the X2-U is the user plane interface between eNode-Bs. It is assumed that there always exists an X2 interface between eNode-Bs that need to communicate with each other, for example, for support of handover.

The specific functions supported by each component and the details about reference points (S1-MME, S1-U, S3, etc.) can be found in [1]. For other nodes in Figure 6.2, the Policy and Charging Rules Function (PCRF) is for policy and charging control, the Home Subscriber Server (HSS) is responsible for the service authorization and user authentication, and the Serving GPRS Support Node (SGSN) is for controlling packet sessions and managing the mobility of the UE for GPRS networks. The topics in this book mainly focus on the E-UTRAN and the LTE radio interface.

6.1.3 Radio Interface Protocols

As in other communication standards, the LTE radio interface is designed based on a layered protocol stack, which can be divided into control plane and user plane protocol stacks and is shown in Figure 6.3. The packet flow in the user plane is shown in Figure 6.4. The LTE radio interface protocol is composed of the following layers:

- **Radio Resource Control (RRC):** The RRC layer performs the control plane functions including paging, maintenance and release of an RRC connection-security handling-mobility management, and QoS management.
- **Packet Data Convergence Protocol (PDCP):** The main functions of the PDCP sublayer include IP packet header compression and decompression based

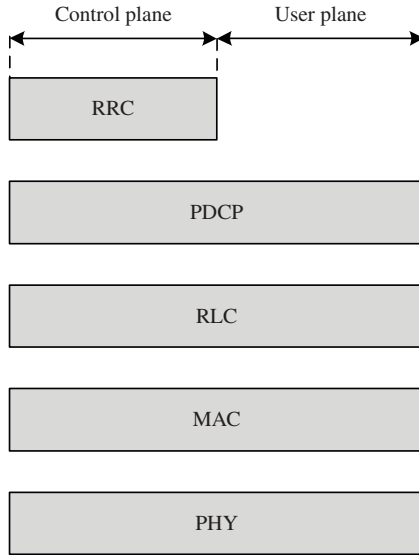


Figure 6.3 The LTE radio interface protocol stack.

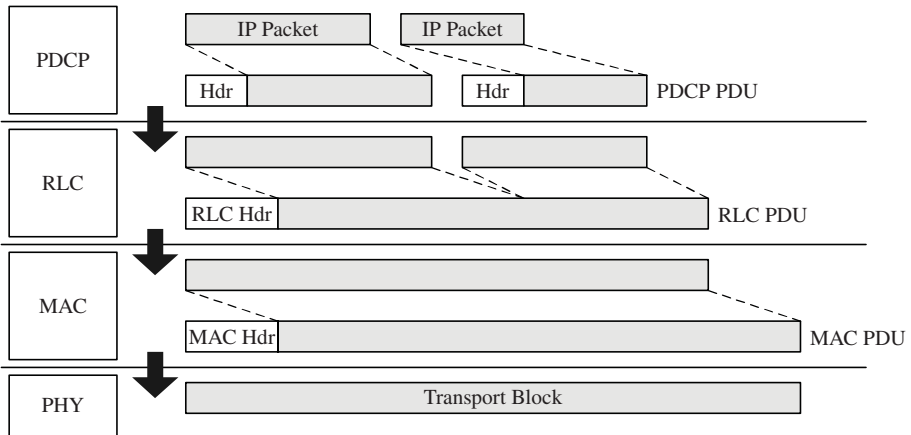


Figure 6.4 The packet flow in the user plane.

on the RObust Header Compression (ROHC) protocol, ciphering of data and signaling, and integrity protection for signaling. There is only one PDCP entity at the eNode-B and the UE per bearer.³

³ A bearer is an IP packet flow with a defined QoS between the PDN GW and the UE. It will be discussed in more detail in Chapter 10.

- **Radio Link Control (RLC):** The main functions of the RLC sublayer are segmentation and concatenation of data units, error correction through the Automatic Repeat reQuest (ARQ) protocol, and in-sequence delivery of packets to the higher layers. It operates in three modes:
 - **The Transparent Mode (TM):** The TM mode is the simplest one, without RLC header addition, data segmentation, or concatenation, and it is used for specific purposes such as random access.
 - **The Unacknowledged Mode (UM):** The UM mode allows the detection of packet loss and provides packet reordering and reassembly, but does not require retransmission of the missing protocol data units (PDUs).
 - **The Acknowledged Mode (AM):** The AM mode is the most complex one, and it is configured to request retransmission of the missing PDUs in addition to the features supported by the UM mode.

There is only one RLC entity at the eNode-B and the UE per bearer.

- **Medium Access Control (MAC):** The main functions of the MAC sublayer include error correction through the Hybrid-ARQ (H-ARQ) mechanism, mapping between logical channels and transport channels, multiplexing/demultiplexing of RLC PDUs on to transport blocks, priority handling between logical channels of one UE, and priority handling between UEs by means of dynamic scheduling. The MAC sublayer is also responsible for transport format selection of scheduled UEs, which includes selection of modulation format, code rate, MIMO rank, and power level. There is only one MAC entity at the eNode-B and one MAC entity at the UE.
- **Physical Layer (PHY):** The main function of PHY is the actual transmission and reception of data in forms of transport blocks. The PHY is also responsible for various control mechanisms such as signaling of H-ARQ feedback, signaling of scheduled allocations, and channel measurements.

In Chapter 7 to Chapter 9, we focus on the PHY layer, also referred to as layer 1 of the Open Systems Interconnection (OSI) reference model. Higher layer processing is described in Chapter 10.

6.2 Hierarchical Channel Structure of LTE

To efficiently support various QoS classes of services, LTE adopts a hierarchical channel structure. There are three different channel types defined in LTE—logical channels, transport channels, and physical channels, each associated with a service access point (SAP) between different layers. These channels are used by the lower layers of the protocol stack to provide services to the higher layers. The radio interface protocol architecture and the SAPs between different layers are shown in Figure 6.5. Logical channels provide services at the SAP between MAC and RLC layers, while transport channels provide services at the SAP between MAC and PHY layers. Physical channels are the actual implementation of transport channels over the radio interface.

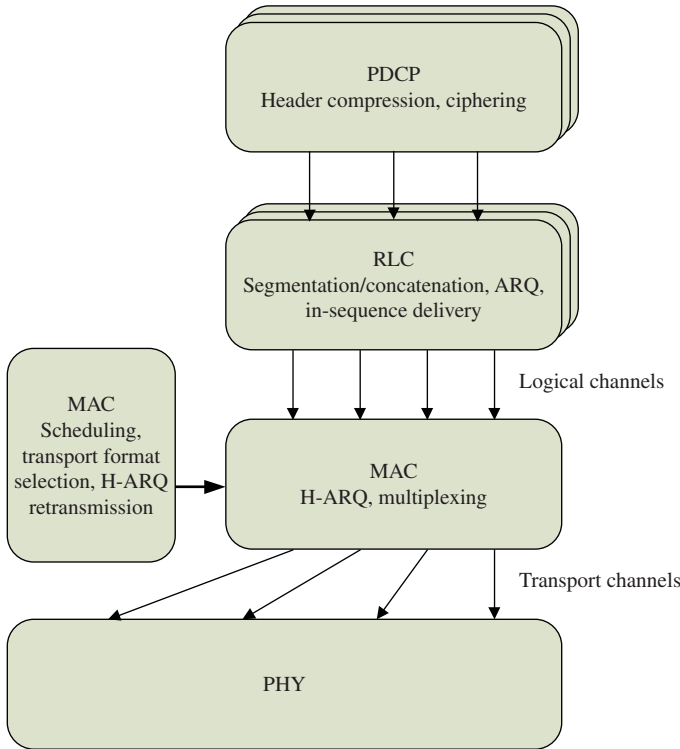


Figure 6.5 The radio interface protocol architecture and the SAPs between different layers.

The channels defined in LTE follow a similar hierarchical structure to UTRA/HSPA. However, in the case of LTE, the transport and logical channel structures are much more simplified and fewer in number compared to UTRA/HSPA. Unlike UTRA/HSPA, LTE is based entirely on shared and broadcast channels and contains no dedicated channels carrying data to specific UEs. This improves the efficiency of the radio interface and can support dynamic resource allocation between different UEs depending on their traffic/QoS requirements and their respective channel conditions. In this section, we describe in detail the various logical, transport, and physical channels that are defined in LTE. The description of different channel types and the channel mapping between different protocol layers provides an intuitive manner to understand the data flow of different services in LTE, which builds the foundation to understand the detail processing procedures in later chapters.

6.2.1 Logical Channels: What to Transmit

Logical channels are used by the MAC to provide services to the RLC. Each logical channel is defined based on the type of information it carries. In LTE, there are two categories of logical channels depending on the service they provide: *logical control channels* and *logical traffic channels*.

The logical control channels, which are used to transfer control plane information, include the following types:

- **Broadcast Control Channel (BCCH):** A downlink common channel used to broadcast system control information to the mobile terminals in the cell, including downlink system bandwidth, antenna configuration, and reference signal power. Due to the large amount of information carried on the BCCH, it is mapped to two different transport channels: the Broadcast Channel (BCH) and the Downlink Shared Channel (DL-SCH).
- **Multicast Control Channel (MCCH):** A point-to-multipoint downlink channel used for transmitting control information to UEs in the cell. It is only used by UEs that receive multicast/broadcast services.
- **Paging Control Channel (PCCH):** A downlink channel that transfers paging information to registered UEs in the cell, for example, in case of a mobile-terminated communication session. The paging process is discussed in Chapter 10.
- **Common Control Channel (CCCH):** A bi-directional channel for transmitting control information between the network and UEs when no RRC connection is available, implying the UE is not attached to the network such as in the idle state. Most commonly the CCCH is used during the random access procedure.
- **Dedicated Control Channel (DCCH):** A point-to-point, bi-directional channel that transmits dedicated control information between a UE and the network. This channel is used when the RRC connection is available, that is, the UE is attached to the network.

The logical traffic channels, which are to transfer user plane information, include:

- **Dedicated Traffic Channel (DTCH):** A point-to-point, bi-directional channel used between a given UE and the network. It can exist in both uplink and downlink.
- **Multicast Traffic Channel (MTCH):** A unidirectional, point-to-multipoint data channel that transmits traffic data from the network to UEs. It is associated with the multicast/broadcast service.

6.2.2 Transport Channels: How to Transmit

The transport channels are used by the PHY to offer services to the MAC. A transport channel is basically characterized by how and with what characteristics data is transferred over the radio interface, that is, the channel coding scheme, the modulation scheme, and antenna mapping. Compared to UTRA/HSPA, the number of transport channels in LTE is reduced since no dedicated channels are present.

LTE defines two MAC entities: one in the UE and one in the E-UTRAN, which handle the following downlink/uplink transport channels.

Downlink Transport Channels

- **Downlink Shared Channel (DL-SCH):** Used for transmitting the downlink data, including both control and traffic data, and thus it is associated with both logical control and logical traffic channels. It supports H-ARQ, dynamic link adaptation, dynamic and semi-persistent resource allocation, UE discontinuous reception, and multicast/broadcast transmission. The concept of shared channel transmission originates from HSDPA, which uses the *High-Speed Downlink Shared Channel* (HS-DSCH) to multiplex traffic and control information among different UEs. By sharing the radio resource among different UEs the DL-SCH is able to maximize the throughput by allocating the resources to the optimum UEs. The processing of the DL-SCH is described in Section 7.2.
- **Broadcast Channel (BCH):** A downlink channel associated with the BCCH logical channel and is used to broadcast system information over the entire coverage area of the cell. It has a fixed transport format defined by the specifications. The processing of the BCH will be described in Section 7.4.
- **Multicast Channel (MCH):** Associated with MCCH and MTCH logical channels for the multicast/broadcast service. It supports *Multicast/Broadcast Single Frequency Network* (MBSFN) transmission, which transmits the same information on the same radio resource from multiple synchronized base stations to multiple UEs. The processing of the MCH is described in Section 7.5.
- **Paging Channel (PCH):** Associated with the PCCH logical channel. It is mapped to dynamically allocated physical resources, and is required for broadcast over the entire cell coverage area. It is transmitted on the Physical Downlink Shared Channel (PDSCH), and supports UE discontinuous reception.

Uplink Transport Channels

- **Uplink Shared Channel (UL-SCH):** The uplink counterpart of the DL-SCH. It can be associated to CCCH, DCCH, and DTCH logical channels. It supports H-ARQ, dynamic link adaptation, and dynamic and semi-persistent resource allocation. The processing of the UL-SCH is described in Section 8.2.
- **Random Access Channel (RACH):** A specific transport channel that is not mapped to any logical channel. It transmits relatively small amounts of data for initial access or, in the case of RRC, state changes. The processing of the RACH is described in Section 8.5, while the random access procedure is described in Section 9.9.

The data on each transport channel is organized into *transport blocks*, and the transmission time of each transport block, also called Transmission Time Interval (TTI), is 1 ms in LTE. TTI is also the minimum interval for link adaptation and scheduling decision. Without spatial multiplexing, at most one transport block is transmitted to a UE in each TTI; with spatial multiplexing, up to two transport blocks can be transmitted in each TTI to a UE.

Besides transport channels, there are different types of control information defined in the MAC layer, which are important for various physical layer procedures. The defined control information includes

- **Downlink Control Information (DCI):** It carries information related to downlink/uplink scheduling assignment, modulation and coding scheme, and Transmit Power Control (TPC) command, and is sent over the Physical Downlink Control Channel (PDCCH). The DCI supports 10 different formats, listed in Table 6.1. Among them, Format 0 is for signaling uplink transmission allocation, Format 3 and 3A are for TPC, and the remaining formats are for signaling downlink transmission allocation. The detail content of each format can be found in [7], some of which is discussed in Section 7.3.
- **Control Format Indicator (CFI):** It indicates how many symbols the DCI spans in that subframe. It takes values $CFI = 1, 2, \text{ or } 3$, and is sent over the Physical Control Format Indicator Channel (PCFICH).
- **H-ARQ Indicator (HI):** It carries H-ARQ acknowledgment in response to uplink transmissions, and is sent over the Physical Hybrid ARQ Indicator Channel (PHICH). $HI = 1$ for a positive acknowledgment (ACK) and $HI = 0$ for a negative acknowledgment (NAK).

Table 6.1 DCI Formats

Format	Carried Information
Format 0	Uplink scheduling assignment
Format 1	Downlink scheduling for one codeword
Format 1A	Compact downlink scheduling for one codeword and random access procedure
Format 1B	Compact downlink scheduling for one codeword with precoding information
Format 1C	Very compact downlink scheduling for one codeword
Format 1D	Compact downlink scheduling for one codeword with precoding and power offset information
Format 2	Downlink scheduling for UEs configured in closed-loop spatial multiplexing mode
Format 2A	Downlink scheduling for UEs configured in open-loop spatial multiplexing mode
Format 3	TPC commands for PUCCH and PUSCH with 2-bit power adjustments
Format 3A	TPC commands for PUCCH and PUSCH with 1-bit power adjustments

- **Uplink Control Information (UCI):** It is for measurement indication on the downlink transmission, scheduling request of uplink, and the H-ARQ acknowledgment of downlink transmissions. The UCI can be transmitted either on the Physical Uplink Control Channel (PUCCH) or the Physical Uplink Shared Channel (PUSCH). The detail transmission format is discussed in Section 8.3.

6.2.3 Physical Channels: Actual Transmission

Each physical channel corresponds to a set of resource elements in the time-frequency grid that carry information from higher layers. The basic entities that make a physical channel are resource elements and resource blocks. A resource element is a single sub-carrier over one OFDM symbol, and typically this could carry one (or two with spatial multiplexing) modulated symbol(s). A resource block is a collection of resource elements and in the frequency domain this represents the smallest quanta of resources that can be allocated. The details of the time-frequency resource structures for downlink and uplink are described in Section 6.3 and Section 6.4, respectively.

Downlink Physical Channels

- **Physical Downlink Control Channel (PDCCH):** It carries information about the transport format and resource allocation related to the DL-SCH and PCH transport channels, and the H-ARQ information related to the DL-SCH. It also informs the UE about the transport format, resource allocation, and H-ARQ information related to UL-SCH. It is mapped from the DCI transport channel.
- **Physical Downlink Shared Channel (PDSCH):** This channel carries user data and higher-layer signaling. It is associated to DL-SCH and PCH.
- **Physical Broadcast Channel (PBCH):** It corresponds to the BCH transport channel and carries system information.
- **Physical Multicast Channel (PMCH):** It carries multicast/broadcast information for the MBMS service.
- **Physical Hybrid-ARQ Indicator Channel (PHICH):** This channel carries H-ARQ ACK/NAKs associated with uplink data transmissions. It is mapped from the HI transport channel.
- **Physical Control Format Indicator Channel (PCFICH):** It informs the UE about the number of OFDM symbols used for the PDCCH. It is mapped from the CFI transport channel.

Uplink Physical Channels

- **Physical Uplink Control Channel (PUCCH):** It carries uplink control information including Channel Quality Indicators (CQI), ACK/NAKs for H-ARQ in response to downlink transmission, and uplink scheduling requests.
- **Physical Uplink Shared Channel (PUSCH):** It carries user data and higher-layer signaling. It corresponds to the UL-SCH transport channel.

- **Physical Random Access Channel (PRACH):** This channel carries the random access preamble sent by UEs.

Besides physical channels, there are signals embedded in the downlink and uplink physical layer, which do not carry information from higher layers. The physical signals defined in the LTE specifications are

- **Reference signal:** It is defined in both downlink and uplink for channel estimation that enables coherent demodulation and for channel quality measurement to assist user scheduling. There are three different reference signals in the downlink:
 - Cell-specific reference signals, associated with non-MBSFN transmission
 - MBSFN reference signals, associated with MBSFN transmission
 - UE-specific reference signals

There are two types of uplink reference signals:

- Demodulation reference signal, associated with transmission of PUSCH or PUCCH
- Sounding reference signal, to support uplink channel-dependent scheduling

The processing of reference signals in the downlink and uplink are treated in Section 7.6.1 and Section 8.4, respectively.

- **Synchronization signal:** It is split into a primary and a secondary synchronization signal, and is only defined in the downlink to enable acquisition of symbol timing and the precise frequency of the downlink signal. It is discussed further in Section 7.6.2.

6.2.4 Channel Mapping

From the description of different channel types, we see that there exists a good correlation based on the purpose and the content between channels in different layers. This requires a mapping between the logical channels and transport channels at the MAC SAP and a mapping between transport channels and physical channels at the PHY SAP. Such channel mapping is not arbitrary, and the allowed mapping between different channel types is shown in Figure 6.6,⁴ while the mapping between control information and physical channels is shown in Figure 6.7. It is possible for multiple channels mapped to a single channel, for example, different logical control channels and logical traffic channels are mapped to the DL-SCH transport channel. The channel mapping in Figures 6.6 and 6.7 will reappear in different sections in Chapters 7 and 8 when we discuss downlink and uplink transport channel processing.

⁴ The mapping of multicast-related channels, that is, MCCH, MTCH, MCH, and PMCH, is not specified in Release 8 but in Release 9.

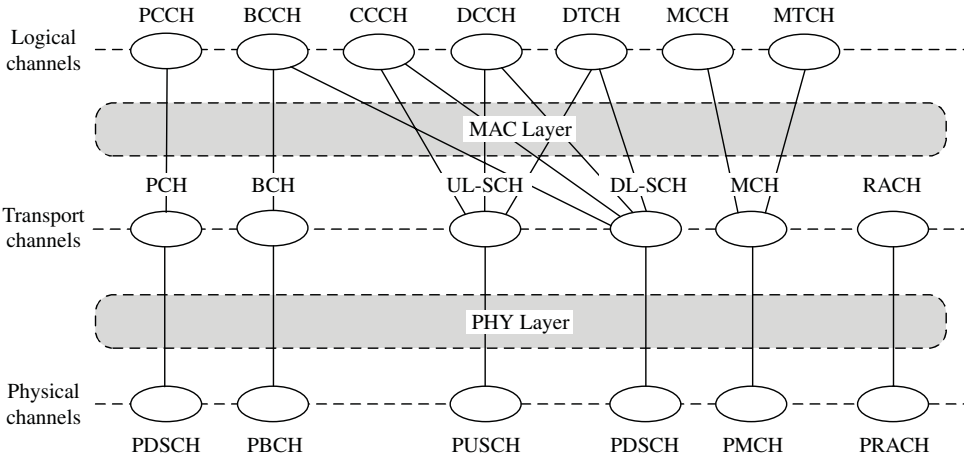


Figure 6.6 Mapping between different channel types.

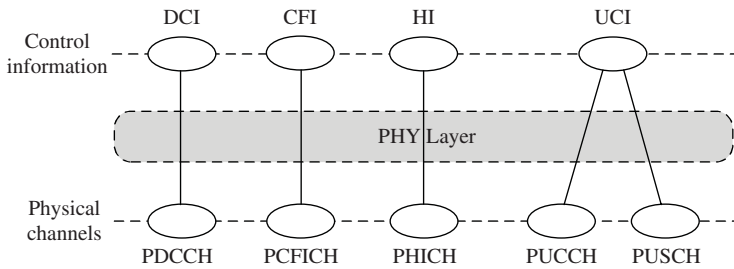


Figure 6.7 Mapping of control information to physical channels.

6.3 Downlink OFDMA Radio Resources

In LTE, the downlink and uplink use different transmission schemes due to different considerations. In this and the next section, we describe downlink and uplink radio transmission schemes, respectively. In the downlink, a scalable OFDM transmission/multiaccess technique is used that allows for high spectrum efficiency by utilizing multiuser diversity in a frequency selective channel. On the other hand, a scalable SC-FDMA transmission/multiaccess technique is used in the uplink since this reduces the peak-to-average power ratio (PAPR) of the transmitted signal.

The downlink transmission is based on OFDM with a cyclic prefix (CP), which was described in Chapter 3 along with the associated multiple access scheme described in Chapter 4. We summarize some key advantages of OFDM that motivate using it in the LTE downlink:

- As shown in Chapter 3, OFDM is efficient in combating the frequency-selective fading channel with a simple frequency-domain equalizer, which makes it a suitable technique for wireless broadband systems such as LTE.

- As shown in Chapter 4, it is possible to exploit frequency-selective scheduling with OFDM-based multiple access (OFDMA), while HSPA only schedules in the time domain. This can make a big difference especially in slow time-varying channels.
- The transceiver structure of OFDM with FFT/IFFT enables scalable bandwidth operation with a low complexity, which is one of the major objectives of LTE.
- As each subcarrier becomes a flat fading channel, compared to single-carrier transmission OFDM makes it much easier to support multiantenna transmission, which is a key technique to enhance the spectrum efficiency.
- OFDM enables multicast/broadcast services on a synchronized single frequency network, that is, MBSFN, as it treats signals from different base stations as propagating through a multipath channel and can efficiently combine them.

The multiple access in the downlink is based on OFDMA. In each TTI, a scheduling decision is made where each scheduled UE is assigned a certain amount of radio resources in the time and frequency domain. The radio resources allocated to different UEs are orthogonal to each other, which means there is no intra-cell interference. In the remaining part of this section, we describe the frame structure and the radio resource block structure in the downlink, as well as the basic principles of resource allocation and the supported MIMO modes.

6.3.1 Frame Structure

Before going into details about the resource block structure for the downlink, we first describe the frame structure in the time domain, which is a common element shared by both downlink and uplink.

In LTE specifications, the size of elements in the time domain is expressed as a number of time units $T_s = 1/(15000 \times 2048)$ seconds. As the normal subcarrier spacing is defined to be $\Delta f = 15\text{kHz}$, T_s can be regarded as the sampling time of an FFT-based OFDM transmitter/receiver implementation with FFT size $N_{\text{FFT}} = 2048$. Note that this is just for notation purpose, as different FFT sizes are supported depending on the transmission bandwidths. A set of parameters for typical transmission bandwidths for LTE in the downlink is shown in Table 6.2, where the subcarrier spacing is $\Delta f = 15\text{kHz}$. The FFT size increases with the transmission bandwidth, ranging from 128 to 2048. With $\Delta f = 15\text{kHz}$, the sampling frequency, which equals $\Delta f \times N_{\text{FFT}}$, is a multiple or sub-multiple of the UTRA/HSPA chip rate of 3.84MHz. In this way, multimode UTRA/HSPA/LTE terminals can be implemented with a single clock circuitry. In addition to the 15kHz subcarrier spacing, a *reduced subcarrier spacing* of 7.5kHz is defined for MBSFN cells, which provides a larger OFDM symbol duration that is able to combat the large delay spread associated with the MBSFN transmission. Unless otherwise stated, we will assume $\Delta f = 15\text{kHz}$ in the following discussion.

In the time domain, the downlink and uplink multiple TTIs are organized into radio frames with duration $T_f = 307200 \cdot T_s = 10$ ms. For flexibility, LTE supports both FDD

Table 6.2 Typical Parameters for Downlink Transmission

Transmission bandwidth [MHz]	1.4	3	5	10	15	20
Occupied bandwidth [MHz]	1.08	2.7	4.5	9.0	13.5	18.0
Guardband [MHz]	0.32	0.3	0.5	1.0	1.5	2.0
Guardband, % of total	23	10	10	10	10	10
Sampling frequency [MHz]	1.92 $1/2 \times 3.84$	3.84	7.68 2×3.84	15.36 4×3.84	23.04 6×3.84	30.72 8×3.84
FFT size	128	256	512	1024	1536	2048
Number of occupied subcarriers	72	180	300	600	900	1200
Number of resource blocks	6	15	25	50	75	100
Number of CP samples (normal)	9×6 10×1	18×6 20×1	36×6 40×1	72×6 80×1	108×6 120×1	144×6 160×1
Number of CP samples (extended)	32	64	128	256	384	512

and TDD modes.⁵ Most of the design parameters are common to FDD and TDD in order to reduce the terminal complexity and maximize reuse between the designs of FDD and TDD systems. Accordingly, LTE supports two kinds of frame structures: frame structure type 1 for the FDD mode and frame structure type 2 for the TDD mode.

Frame Structure Type 1

Frame structure type 1 is applicable to both full duplex and half duplex FDD. There are three different kinds of units specified for this frame structure, illustrated in Figure 6.8. The smallest one is called a *slot*, which is of length $T_{slot} = 15360 \cdot T_s = 0.5$ ms. Two consecutive slots are defined as a *subframe* of length 1 ms, and 20 slots, numbered from 0 to 19, constitute a *radio frame* of 10 ms. Channel-dependent scheduling and link adaptation operate on a subframe level. Therefore, the subframe duration corresponds to the minimum downlink TTI, which is of 1 ms duration, compared to a 2 ms TTI for the HSPA and a minimum 10 ms TTI for the UMTS. A shorter TTI is for fast link adaptation and is able to reduce delay and better exploit the time-varying channel through channel-dependent scheduling.

Each slot consists of a number of OFDM symbols including CPs. As shown in Chapter 3, CP is a kind of guard interval to combat inter-OFDM-symbol interference, which should be larger than the channel delay spread. Therefore, the length of CP depends on the environment where the network operates, and it should not be too large as it brings a bandwidth and power penalty. With a subcarrier spacing $\Delta f = 15$ kHz, the OFDM symbol time is $1/\Delta f \approx 66.7 \mu s$. As shown in Figure 6.8, LTE defines two different

⁵ The LTE TDD mode, also referred to as TD-LTE, provides the long-term evolution path for TD-SCDMA-based networks.

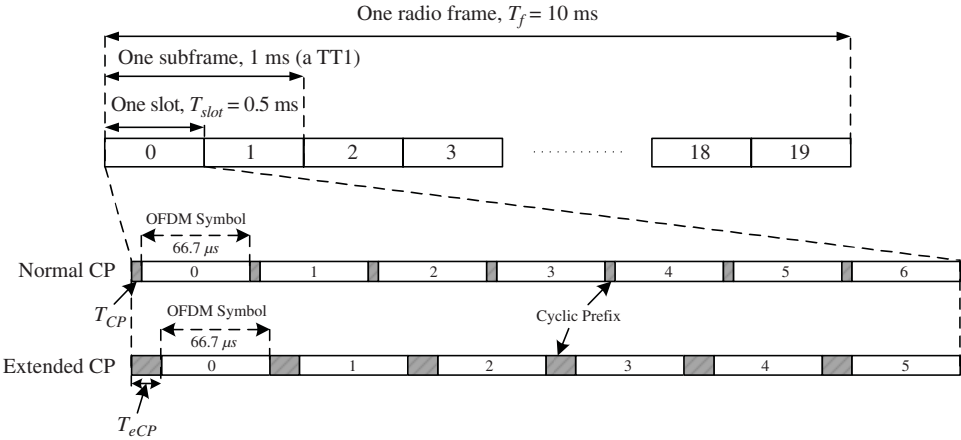


Figure 6.8 Frame structure type 1. For the normal CP, $T_{CP} = 160 \cdot T_s \approx 5.2 \mu s$ for the first OFDM symbol, and $T_{CP} = 144 \cdot T_s \approx 4.7 \mu s$ for the remaining OFDM symbols, which together fill the entire slot of 0.5 ms. For the extended CP, $T_{eCP} = 512 \cdot T_s \approx 16.7 \mu s$.

CP lengths: a *normal CP* and an *extended CP*, corresponding to seven and six OFDM symbols per slot, respectively. The extended CP is for multicell multicast/broadcast and very-large-cell scenarios with large delay spread at a price of bandwidth efficiency, with length $T_{eCP} = 512 \cdot T_s \approx 16.7 \mu s$. The normal CP is suitable for urban environment and high data rate applications. Note that the normal CP lengths are different for the first ($T_{CP} = 160 \cdot T_s \approx 5.2 \mu s$) and subsequent OFDM symbols ($T_{CP} = 144 \cdot T_s \approx 4.7 \mu s$), which is to fill the entire slot of 0.5 ms. The numbers of CP samples for different bandwidths are shown in Table 6.2. For example, with 10MHz bandwidth, the sampling time is $1/(15000 \times 1024)$ sec and the number of CP samples for the extended CP is 256, which provides the required CP length of $256/(15000 \times 1024) \approx 1.67 \mu s$. In case of 7.5kHz subcarrier spacing, there is only a single CP length, corresponding to 3 OFDM symbols per slot.

For FDD, uplink and downlink transmissions are separated in the frequency domain, each with 10 subframes. In half-duplex FDD operation, the UE cannot transmit and receive at the same time while there are no such restrictions in full-duplex FDD. However, full-duplex FDD terminals need high quality and expensive RF duplex-filters to separate uplink and downlink channels, while half-duplex FDD allows hardware sharing between the uplink and downlink, which offers a cost saving at the expense of reducing data rates by half. Half-duplex FDD UEs are also considered a good solution if the duplex separation between the uplink and downlink transmissions is relatively small. In such cases, the half-duplex FDD is the preferable approach to mitigate the cross-interference between the transmit and receive chains.

Frame Structure Type 2

Frame structure type 2 is applicable to the TDD mode. It is designed for coexistence with legacy systems such as the 3GPP TD-SCDMA-based standard. As shown in Figure 6.9, each radio frame of frame structure type 2 is of length $T_f = 30720 \cdot T_s = 10$ ms, which

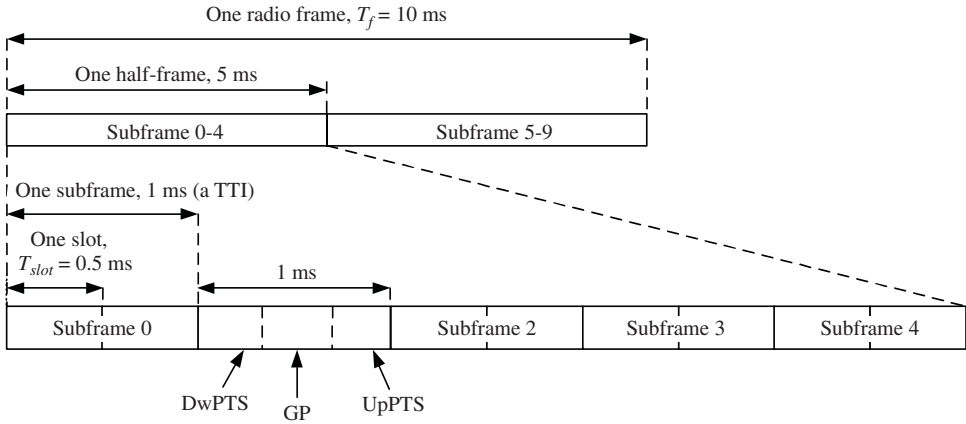


Figure 6.9 Frame structure type 2.

consists of two half-frames of length 5 ms each. Each half-frame is divided into five subframes with 1 ms duration. There are special subframes, which consist of three fields: Downlink Pilot TimeSlot (DwPTS), Guard Period (GP), and Uplink Pilot TimeSlot (UpPTS). These fields are already defined in TD-SCDMA and are maintained in the LTE TDD mode to provide sufficiently large guard periods for the equipment to switch between transmission and reception.

- **The DwPTS field:** This is the downlink part of the special subframe, and can be regarded as an ordinary but shorter downlink subframe for downlink data transmission. Its length can be varied from three up to twelve OFDM symbols.
- **The UpPTS field:** This is the uplink part of the special subframe, and has a short duration with one or two OFDM symbols. It can be used for transmission of uplink sounding reference signals and random access preambles.
- **The GP field:** The remaining symbols in the special subframe that have not been allocated to DwPTS or UpPTS are allocated to the GP field, which is used to provide the guard period for the downlink-to-uplink and the uplink-to-downlink switch.

The total length of these three special fields has a constraint of 1 ms. With the DwPTS and UpPTS durations mentioned above, LTE supports a guard period ranging from two to ten OFDM symbols, sufficient for cell size up to and beyond 100 km. All other subframes are defined as two slots, each with length $T_{slot} = 0.5$ ms.

Figure 6.9 only shows the detail structure of the first half-frame. The second half-frame has the similar structure, which depends on the uplink-downlink configuration. Seven uplink-downlink configurations with either 5 ms or 10 ms downlink-to-uplink switch-point periodicity are supported, as illustrated in Table 6.3, where “D” and “U” denote subframes reserved for downlink and uplink, respectively, and “S” denotes the special

Table 6.3 Uplink-Downlink Configurations for the LTE TDD Mode

Uplink-Downlink Configuration	Downlink-to-Uplink Switch-Point Periodicity	Subframe Number									
		0	1	2	3	4	5	6	7	8	9
0	5 ms	D	S	U	U	U	D	S	U	U	U
1	5 ms	D	S	U	U	D	D	S	U	U	D
2	5 ms	D	S	U	D	D	D	S	U	D	D
3	10 ms	D	S	U	U	U	D	D	D	D	D
4	10 ms	D	S	U	U	D	D	D	D	D	D
5	10 ms	D	S	U	D	D	D	D	D	D	D
6	5 ms	D	S	U	U	U	D	S	U	U	D

subframe. In the case of 5 ms switch-point periodicity, the special subframe exists in both half-frames, and the structure of the second half-frame is the same as the first one depicted in Figure 6.9. In the case of 10 ms switch-point periodicity, the special subframe exists in the first half-frame only. Subframes 0, 5, and the field DwPTS are always reserved for downlink transmission, while UpPTS and the subframe immediately following the special subframe are always reserved for uplink transmission.

6.3.2 Physical Resource Blocks for OFDMA

The physical resource in the downlink in each slot is described by a time-frequency grid, called a *resource grid*, as illustrated in Figure 6.10. Such a time-frequency plane representation is a common practice for OFDM systems, which makes it intuitive for radio resource allocation. Each column and each row of the resource grid correspond to one OFDM symbol and one OFDM subcarrier, respectively. The duration of the resource grid in the time domain corresponds to one slot in a radio frame. The smallest time-frequency unit in a resource grid is denoted as a *resource element*. Each resource grid consists of a number of *resource blocks*, which describe the mapping of certain physical channels to resource elements. The detail of these resource units is described as follows.

Resource Grid

The structure of each resource grid is characterized by the following three parameters:

- **The number of downlink resource blocks (N_{RB}^{DL}):** It depends on the transmission bandwidth and shall fulfill $N_{RB}^{min,DL} \leq N_{RB}^{DL} \leq N_{RB}^{max,DL}$, where $N_{RB}^{min,DL} = 6$ and $N_{RB}^{max,DL} = 110$ are for the smallest and largest downlink channel bandwidth, respectively. The values of N_{RB}^{DL} for several current specified bandwidths are listed in Table 6.2.
- **The number of subcarriers in each resource block (N_{sc}^{RB}):** It depends on the subcarrier spacing Δf , satisfying $N_{sc}^{RB} \Delta f = 180\text{kHz}$, that is, each resource block is

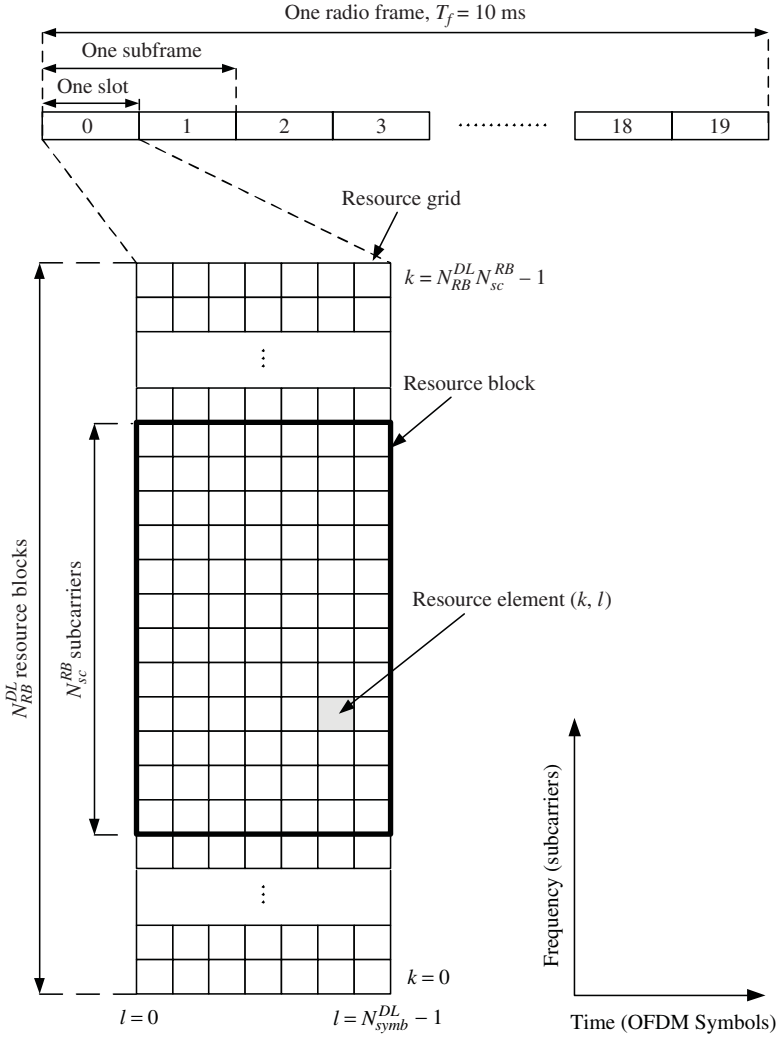


Figure 6.10 The structure of the downlink resource grid.

of 180kHz wide in the frequency domain. The values of N_{sc}^{RB} for different subcarrier spacings are shown in Table 6.4. There are a total of $N_{RB}^{DL} \times N_{sc}^{RB}$ subcarriers in each resource grid. For downlink transmission, the DC subcarrier is not used as it may be subject to a too high level of interference.

- **The number of OFDM symbols in each block (N_{symp}^{DL}):** It depends on both the CP length and the subcarrier spacing, specified in Table 6.4.

Therefore, each downlink resource grid has $N_{RB}^{DL} \times N_{sc}^{RB} \times N_{symp}^{DL}$ resource elements. For example, with 10MHz bandwidth, $\Delta f = 15$ kHz, and normal CP, we get $N_{RB}^{DL} = 50$ from

Table 6.4 Physical Resource Block Parameters for the Downlink

Configuration	N_{sc}^{RB}	N_{symp}^{DL}
Normal CP $\Delta f = 15\text{kHz}$	12	7
Extended CP $\Delta f = 15\text{kHz}$	12	6
$\Delta f = 7.5\text{kHz}$	24	3

Table 6.2, $N_{RB}^{sc} = 12$ and $N_{symp}^{DL} = 7$ from Table 6.4, so there are $50 \times 12 \times 7 = 4200$ resource elements in the downlink resource grid.

In case of multiantenna transmission, there is one resource grid defined per antenna port. An antenna port is defined by its associated reference signal, which may not correspond to a physical antenna. The set of antenna ports supported depends on the reference signal configuration in the cell. As discussed in Section 6.2.3, there are three different reference signals defined in the downlink, and the associated antenna ports are as follows:

- Cell-specific reference signals support a configuration of 1, 2, or 4 antenna ports and the antenna port number p shall fulfill $p = 0$, $p \in \{0, 1\}$, and $p \in \{0, 1, 2, 3\}$, respectively.
- MBSFN reference signals are transmitted on antenna port $p = 4$.
- UE-specific reference signals are transmitted on antenna port $p = 5$.

We will talk more about antenna ports when discussing MIMO transmission in the downlink in Section 7.2.2.

Resource Element

Each resource element in the resource grid is uniquely identified by the index pair (k, l) in a slot, where $k = 0, 1, \dots, N_{RB}^{DL} N_{sc}^{RB} - 1$ and $l = 0, 1, \dots, N_{symp}^{DL} - 1$ are indices in the frequency and time domains, respectively. The size of each resource element depends on the subcarrier spacing Δf and the CP length.

Resource Block

The resource block is the basic element for radio resource allocation. The minimum size of radio resource that can be allocated is the minimum TTI in the time domain, that is, one subframe of 1 ms, corresponding to two resource blocks. The size of each resource block is the same for all bandwidths, which is 180kHz in the frequency domain. There are two kinds of resource blocks defined for LTE: physical and virtual resource blocks, which are defined for different resource allocation schemes and are specified in the following section.

6.3.3 Resource Allocation

Resource allocation's role is to dynamically assign available time-frequency resource blocks to different UEs in an efficient way to provide good system performance. In LTE,

channel-dependent scheduling is supported, and transmission is based on the shared channel structure where the radio resource is shared among different UEs. Therefore, with resource allocation techniques described in Chapter 4, *multiuser diversity* can be exploited by assigning resource blocks to the UEs with favorable channel qualities. Moreover, resource allocation in LTE is able to exploit the channel variations in both the time and frequency domain, which provides higher multiuser diversity gain than HSPA that can only exploit the time-domain variation. Given a wide bandwidth in LTE, this property is beneficial especially for slow-time varying channels, such as in the scenario with low mobility, where taking advantage of channel selectivity in the time domain is difficult.

With OFDMA, the downlink resource allocation is characterized by the fact that each scheduled UE occupies a number of resource blocks while each resource block is assigned exclusively to one UE at any time. Physical resource blocks (PRBs) and virtual resource blocks (VRBs) are defined to support different kinds of resource allocation types. The VRB is introduced to support both block-wise transmission (localized) and transmission on non-consecutive subcarriers (distributed) as a means to maximize frequency diversity. The LTE downlink supports three resource allocation types: type 0, 1, and 2 [8]. The downlink scheduling is performed at the eNode-B based on the channel quality information fed back from UEs, and then the downlink resource assignment information is sent to UEs on the PDCCH channel.

A PRB is defined as N_{symb}^{DL} consecutive OFDM symbols in the time domain and N_{sc}^{RB} consecutive subcarriers in the frequency domain, as demonstrated in Figure 6.10. Therefore, each PRB corresponds to one slot in the time domain (0.5 ms) and 180kHz in the frequency domain. PRBs are numbered from 0 to $N_{RB}^{DL} - 1$ in the frequency domain. The PRB number n_{PRB} of a resource element (k, l) in a slot is given by:

$$n_{PRB} = \left\lfloor \frac{k}{N_{sc}^{RB}} \right\rfloor.$$

The PRB is to support resource allocations of type 0 and type 1, which are defined for the DCI format 1, 2, and 2A.

- In **type 0 resource allocations**, several consecutive PRBs constitute a resource block group (RBG), and the resource allocation is done in units of RBGs. Therefore, a bitmap indicating the RBG is sufficient to carry the resource assignment. The allocated RBGs to a certain UE do not need to be adjacent to each other, which provides frequency diversity. The RBG size P , that is, the number of PRBs in each RBG, depends on the bandwidth and is specified in Table 6.5. An example of type 0

Table 6.5 Resource Allocation RBG Size vs. Downlink System Bandwidth

Downlink Resource Blocks (N_{RB}^{DL})	RBG Size (P)
≤ 10	1
11 – 26	2
27 – 63	3
64 – 110	4

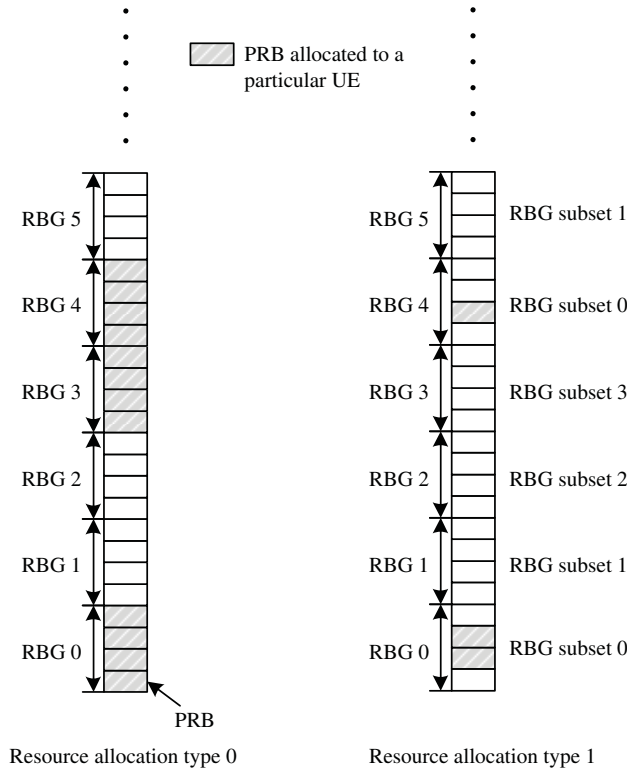


Figure 6.11 Examples of resource allocation type 0 and type 1, where the RBG size $P = 4$.

resource allocation is shown in Figure 6.11, where $P = 4$ and RBGs 0, 3, 4, . . . , are allocated to a particular UE.

- In **type 1 resource allocations**, all the RBGs are grouped into a number of RBG subsets, and certain PRBs inside a selected RBG subset are allocated to the UE. There are a total of P RBG subsets, where P is the RBG size. An RBG subset p , where $0 \leq p < P$, consists of every P -th RBG starting from RBG p . Therefore, the resource assignment information consists of three fields: the first field indicates the selected RBG subset, the second field indicates whether an offset is applied, and the third field contains the bitmap indicating PRBs inside the selected RBG subset. This type of resource allocation is more flexible and is able to provide higher frequency diversity, but it also requires a larger overhead. An example of type 1 resource allocation is shown in Figure 6.11, where $P = 4$ and the RBG subset 0 is selected for the given UE.

In **type 2 resource allocations** that are defined for the DCI format 1A, 1B, 1C, and 1D, PRBs are not directly allocated. Instead, VRBs are allocated, which are then mapped onto PRBs. A VRB is of the same size as a PRB. There are two types of VRBs: VRBs of the localized type and VRBs of the distributed type.

For each type of VRB, a pair of VRBs over two slots in a subframe are assigned together with a single VRB number, n_{VRB} . VRBs of the localized type are mapped directly to physical resource blocks such that the VRB number n_{VRB} corresponds to the PRB number $n_{PRB} = n_{VRB}$. For VRBs of the distributed type, the VRB numbers are mapped to PRB numbers according to the rule specified in [6].

For resource allocations of type 2, the resource assignment information indicates a set of contiguously allocated localized VRBs or distributed VRBs. A one-bit flag indicates whether localized VRBs or distributed VRBs are assigned.

Details about the downlink resource allocation can be found in [8]. The feedback for channel quality information and the related signaling is discussed in Chapter 9.

6.3.4 Supported MIMO Modes

Multiantenna transmission and reception (MIMO), as described in Chapter 5, is a physical layer technique that can improve both the reliability and throughput of the communications over wireless channels. It is considered a key component of the LTE physical layer from the start. The baseline antenna configuration in LTE is two transmit antennas at the cell site and two receive antennas at the UE. The higher-order downlink MIMO is also supported with up to four transmit and four receive antennas.

The downlink transmission supports both single-user MIMO (SU-MIMO) and multiuser MIMO (MU-MIMO). For SU-MIMO, one or multiple data streams are transmitted to a single UE through space-time processing; for MU-MIMO, modulation data streams are transmitted to different UEs using the same time-frequency resource. The supported SU-MIMO modes are listed as follows:

- Transmit diversity with space frequency block codes (SFBC)
- Open-loop spatial multiplexing supporting four data streams
- Closed-loop spatial multiplexing, with closed-loop precoding as a special case when channel rank = 1
- Conventional direction of arrival (DOA)-based beamforming

The supported MIMO mode is restricted by the UE capability. The PDSCH physical channel supports all the MIMO modes, while other physical channels support transmit diversity except PMCH, which only supports single-antenna-port transmission. The details about MIMO transmission on each downlink physical channel are provided in Chapter 7, while the feedback to assist MIMO transmission is discussed in Chapter 9.

6.4 Uplink SC-FDMA Radio Resources

For the LTE uplink transmission, SC-FDMA with a CP is adopted. As discussed in Chapter 4, SC-FDMA possesses most of the merits of OFDM while enjoying a lower PAPR. A lower PAPR is highly desirable in the uplink as less expensive power amplifiers are needed at UEs and the coverage is improved. In LTE, the SC-FDMA signal is

generated by the DFT-spread-OFDM. Compared to conventional OFDM, the SC-FDMA receiver has higher complexity, which, however, is not considered to be an issue in the uplink given the powerful computational capability at the base station.

An SC-FDMA transceiver has a similar structure as OFDM, so the parametrization of radio resource in the uplink enjoys similarities to that in the downlink described in Section 6.3. Nevertheless, the uplink transmission has its own properties. Different from the downlink, only localized resource allocation on consecutive subcarriers is allowed in the uplink. In addition, only limited MIMO modes are supported in the uplink. In this section, we focus on the differences in the uplink radio resource from that in the downlink.

6.4.1 Frame Structure

The uplink frame structure is similar to that for the downlink. The difference is that now we talk about *SC-FDMA symbols* and *SC-FDMA subcarriers*. In frame structure type 1, an uplink radio frame consists of 20 slots of 0.5 ms each, and one subframe consists of two slots, as in Figure 6.8. Frame structure type 2 consists of ten subframes, with one or two special subframes including DwPTS, GP, and UpPTS fields, as shown in Figure 6.9. A CP is inserted prior to each SC-FDMA symbol. Each slot carries seven SC-FDMA symbols in the case of normal CP, and six SC-FDMA symbols in the case of extended CP.

6.4.2 Physical Resource Blocks for SC-FDMA

As SC-FDMA can be regarded as conventional OFDM with a DFT-based precoder, the *resource grid* for the uplink is similar to the one for the downlink, illustrated in Figure 6.12, that is, it comprises a number of resource blocks in the time-frequency plane. The number of resource blocks in each resource grid, N_{RB}^{UL} , depends on the uplink transmission bandwidth configured in the cell and should satisfy

$$N_{RB}^{min,UL} \leq N_{RB}^{UL} \leq N_{RB}^{max,UL},$$

where $N_{RB}^{min,UL} = 6$ and $N_{RB}^{max,UL} = 110$ correspond to the smallest and largest uplink bandwidth, respectively. There are $N_{sc}^{RB} \times N_{symp}^{RB}$ resource elements in each resource block. The values of N_{sc}^{RB} and N_{symp}^{UL} for normal and extended CP are given in Table 6.6. There is only one subcarrier spacing supported in the uplink, which is $\Delta f = 15\text{kHz}$. Different from the downlink, the DC subcarrier is used in the uplink, as the DC interference is spread over the modulation symbols due to the DFT-based precoding.

As for the downlink, each *resource element* in the resource grid is uniquely defined by the index pair (k, l) in a slot, where $k = 0, \dots, N_{RB}^{UL} N_{sc}^{RB} - 1$ and $l = 0, \dots, N_{symp}^{UL} - 1$ are the indices in the frequency and time domain, respectively. For the uplink, no antenna port is defined, as only single antenna transmission is supported in the current specifications.

A PRB in the uplink is defined as N_{symp}^{UL} consecutive SC-FDMA symbols in the time domain and N_{sc}^{RB} consecutive subcarriers in the frequency domain, corresponding to one slot in the time domain and 180kHz in the frequency domain. The relation between the

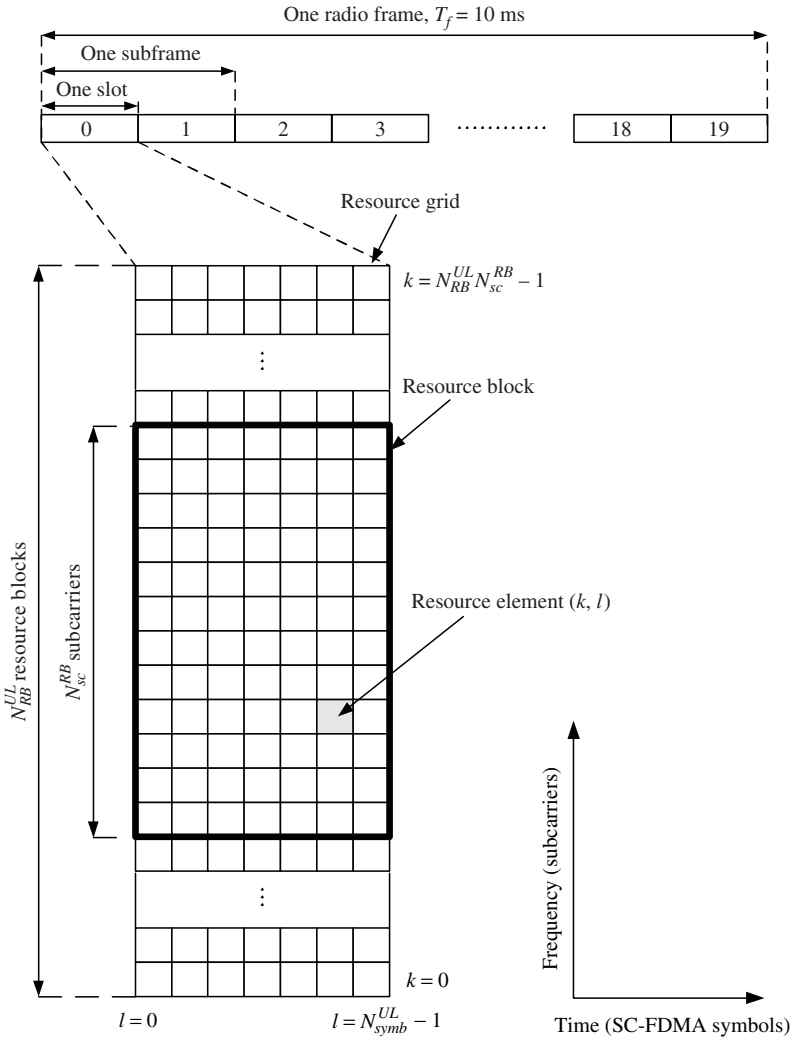


Figure 6.12 The structure of the uplink resource grid.

Table 6.6 Physical Resource Block Parameters for Uplink

Configuration	N_{sc}^{RB}	N_{symb}^{UL}
Normal CP	12	7
Extended CP	12	6

PRB number n_{PRB} in the frequency domain and resource elements (k, l) in a slot is given by:

$$n_{PRB} = \left\lfloor \frac{k}{N_{sc}^{RB}} \right\rfloor.$$

6.4.3 Resource Allocation

Similar to the downlink, shared-channel transmission and channel-dependent scheduling are supported in the uplink. Resource allocation in the uplink is also performed at the eNode-B. Based on the channel quality measured on the uplink sounding reference signals and the scheduling requests sent from UEs, the eNode-B assigns a unique time-frequency resource to a scheduled UE, which achieves orthogonal intra-cell transmission. Such intra-cell orthogonality in the uplink is preserved between UEs by using timing advance such that the transport blocks of different UEs are received synchronously at the eNode-B. This provides significant coverage and capacity gain in the uplink over UMTS, which employs non-orthogonal transmission in the uplink and the performance is limited by inter-channel interference. In general, SC-FDMA is able to support both localized and distributed resource allocation. In the current specification, only localized resource allocation is supported in the uplink, which preserves the single-carrier property and can better exploit the multiuser diversity gain in the frequency domain. Compared to distributed resource allocation, localized resource allocation is less sensitive to frequency offset and also requires fewer reference symbols.

The resource assignment information for the uplink transmission is carried on the PDCCH with DCI format 0, indicating a set of contiguously allocated resource blocks. However, not all integer multiples of one resource block are allowed to be assigned to a UE, which is to simplify the DFT design for the SC-FDMA transceiver. Only factors 2, 3, and 5 are allowed. The frequency hopping is supported to provide frequency diversity, with which the UEs can hop between frequencies within or between the allocated subframes. The resource mapping for different uplink channels is discussed in Chapter 8, and the uplink channel sounding and scheduling signaling is described in Chapter 9.

6.4.4 Supported MIMO Modes

For the MIMO modes supported in the uplink, the terminal complexity and cost are among the major concerns. MU-MIMO is supported, which allocates the same time and frequency resource to two UEs with each transmitting on a single antenna. This is also called Spatial Division Multiple Access (SDMA). The advantage is that only one transmit antenna per UE is required. To separate streams for different UEs, channel state information is required at the eNode-B, which is obtained through uplink reference signals that are orthogonal between UEs. Uplink MU-MIMO also requires power control, as the near-far problem arises when multiple UEs are multiplexed on the same radio resource.

For UEs with two or more transmit antennas, closed-loop adaptive antenna selection transmit diversity shall be supported. For this scenario, each UE only needs one transmit

chain and amplifier. The antenna that provides the best channel to the eNode-B is selected based on the feedback from the eNode-B. The details of MIMO transmission in the uplink are described in Chapter 8.

6.5 Summary and Conclusions

This chapter provided an overview of the LTE radio interface, emphasizing the hierarchical channel structure and the radio resource in both downlink and uplink. The material covered should be adequate for the reader to get the unique characteristics of the LTE physical layer and understand the detailed physical layer procedures in the following chapters.

- LTE is the next step in the evolution of mobile cellular systems, and is a packet-switched network from end to end that is designed with a clean slate approach.
- LTE adopts the hierarchical channel structure from UTRA/HSPA. It simplifies the channel structure and is based totally on the shared channel transmission, which improves the efficiency of the air interface.
- LTE applies OFDMA in the downlink and SC-FDMA in the uplink, both of which have similar radio resource structures in the time-frequency plane. The capability of scheduling in both time and frequency domain provides a higher spectral efficiency in LTE than what is achieved in HSPA. Both localized and distributed resource allocations are supported in the downlink, while only localized resource allocation is supported in the uplink.
- MIMO transmission is a key component of LTE. In current specifications, downlink transmission supports a variety of MIMO modes, while uplink transmission has a limited MIMO support considering cost and complexity.

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Index

Numbers

- 1xRTT/EvDO, 229
- 2-ray approximation for path loss, 50
- 2.5G system, CDMA2000-1X as, 12
- 2G (second generation) digital cellular systems
 - 3G systems vs., 10
 - AMPS, 6
 - CDMA (IS-95), 9–10
 - GSM, 7–9
 - LTE interoperability with, 27, 230
 - overview of, 6–7
- 3G (third generation) broadband wireless systems
 - 3GPP standards, 22–24
 - CDMA 2000 and EV-DO, 11–13
 - GSM vs. CDMA evolution plans, 10
 - HSPA, 14–15
 - LTE interoperability with, 27, 230
 - overview of, 10–11
 - UMTS WCDMA, 13–14
- 3G (third generation) broadband wireless systems, beyond 3G
 - HSPA+, 16–18
 - HSPA+ and WiMAX vs. LTE, 20–22
 - mobile WiMAX, 18–20
 - overview of, 15–16
- 3GPP (Third Generation Partnership Project)
 - channel models, 76
 - LTE flat network architecture, 32–33
 - radio interface standards for IMT-2000, 11
 - releases, 22–23
 - separate network structure for LTE, 227
 - standards evolution, 22–24
- 3GPP2 (Third Generation Partnership Project 2), 11–12
- 4G (fourth generation) systems
 - ITU definition of, 40
 - LTE not meeting requirements for, 39
 - WiMAX and LTE not technically, 16
- 16QAM modulation
 - computing shadowing effects for, 53–54
 - CQI estimation process, 324
 - downlink shared channels, 269
 - HSPA+ enhancements to, 16
 - LTE schemes for, 264
 - in OFDM, 110
 - uplink shared channels, 299
- 64QAM modulation
 - AMC, 86–88
 - CQI estimation process, 324
 - downlink shared channels, 269
 - HSPA+ enhancements to, 16
 - LTE schemes for, 264
 - uplink shared channels, 299

A

- AAS (advanced antenna systems), and WiMAX, 19
- ACK/NAK
 - H-ARQ Indicator transmissions in uplink, 321
 - H-ARQ transmissions in downlink, 320
 - modulation of PUCCH, 307–308
 - multiplexing UCI on PUSCH with UL-SCH data, 304–305
- Acknowledged Mode (AM), RLC, 360, 364
- ACK_SN (Acknowledgement SN) field, STATUS PDU, 366
- adaptive H-ARQ, 292
- ADC (analog-to-digital) converters, 46, 118
- additive white Gaussian noise (AWGN), 80–81, 89
- adjacent subcarrier allocation, 153
- advanced antenna systems (AAS), and WiMAX, 19

- Advanced Mobile Phone Service (AMPS), 1, 5–6
- advanced wireless services (AWS) band, LTE deployment, 35–38
- Alamouti STBC systems
 - diversity, multiuser vs. frequency vs. spatial, 159–160
 - MIMO in LTE downlink, 213–214
 - open-loop transmit diversity, 175–177
 - vs. MRC for coherent BPSK, 181–182
- allocation notification, uplink
 - feedback, 154
- ALOHA, 135
- AM (Acknowledged Mode), RLC, 360, 364
- AM/AM (amplitude modulation-amplitude modulation), and PAR problem, 116
- AM/FM (amplitude modulation-phase modulation), and PAR problem, 116
- AMC (adaptive modulation and coding)
 - computing effect of shadowing, 54
 - HSPDA and, 14
 - mitigating narrowband fading, 86–89
 - tuning controller, 88–89
 - WiMAX link layer and, 19
- AMD (AM Data) PDU, 364–366
- AMD PDU segments
 - and ARQ, 369
 - defined, 364
 - RLC PDU formats, 365–366
- amplitude modulation-amplitude modulation (AM-AM), and PAR problem, 116
- amplitude modulation-phase modulation (AM/FM), and PAR problem, 116
- AMPS (Advanced Mobile Phone Service), 1, 5–6
- analog-to-digital (ADC) converter, 46, 118
- angle of arrival. *See* AoA (angle of arrival)
- angle of departure (AoD), 76, 78
- angle spread (AS), 76, 78
- angular spread, 63, 66–67
- antenna ports
 - cell-specific reference signals, 286–287
 - generating OFDM baseband signals on, 268
 - mapping of resources on, 266–268
 - in MIMO transmission, 265–266, 336
- antennas
 - Advanced Antenna Systems, 19
 - array fashion, 76
 - array gain pattern, 76
 - directional, 59
 - diversity in HSPA+, 17
 - multiple transmission and reception. *See* MIMO (multiantenna transmission and reception)
 - single antenna systems, 159–160
 - uncorrelated, 208–209
- AoA (angle of arrival)
 - beamsteering for interference suppression, 187–189
 - empirical channel models estimating, 76
 - LTE channels for multipath modelling, 78
- AoD (angle of departure), 76, 78
- aperiodic CQI reporting
 - channel feedback and, 207–208
 - defined, 325–326
 - understanding, 331–333
- ARK NAK, triggering ARQ, 369
- ARQ (Automatic Repeat Request) protocol. *See also* H-ARQ (Hybrid-ARQ) protocol
 - correcting link layer errors with, 14
 - defined, 86
 - DMT theory for, 201
 - mitigating narrowband fading, 86
 - procedures in MAC/RLC, 368–369
 - RLC error correction through, 234
 - tuning AMC controller, 89
 - WiMAX link layer supporting, 18
- array gain, 168–169
- array response vector, DOA-based
 - beamsteering, 188
- AS (angle spread), 76, 78
- asynchronous H-ARQ, 292
- AT&T, evolution of, 3, 5–6
- attenuation factor, clipping ratio, 122–123
- autocorrelation function
 - angular spread/coherence distance in, 66–67
 - delay spread/coherence bandwidth in, 64
 - Doppler spread/coherence time in, 65–66
 - for received signals, 71–75
 - two-dimensional, 61–63
 - for wireless channel parameters, 63

- avoidance, ICI, 378–379
- AWGN (additive white Gaussian noise), 80–81, 89
- AWS (advanced wireless services) band, LTE deployment, 35–38
- ## B
- backward compatibility
 LTE-Advanced, 40
 WiMAX developed without constraints of, 20
- band AMC mode, OFDMA, 138, 141
- bandwidth
 CQI feedback design, 322
 cyclic prefix creating loss of, 106–107
 HSPA vs. WiMAX vs. LTE, 20–21
 LTE design requirements, 24–27
 OFDM communications and, 108
 OFDM in LTE, 109–110
 range vs., 49
 WiMAX scalability, 19
- bandwidth part (BP), UE-selected subband CQI report, 328–329
- base-station controller (BSC), 8–9, 13
- base station coordination, networked MIMO, 212–213
- Base Station Subsystem (BSS), GSM, 8–9
- base-station transceiver (BTS), 8–9, 13
- base stations. *See* BS (base stations)
- baseband signals
 OFDM, 268
 SC-FDMA, 297–298
- battery life, HSPA+, 17
- BCCH (Broadcast Control Channel), 236, 284
- BCH (Broadcast Channel), 237
 downlink transport channels, 283
 overview of, 83–84
 uplink transport channels, 237
- beamforming
 DOA-based, 187–189
 interference-aware, 191–192
 LTE multiantennas and, 31
 MIMO in LTE downlink and, 214
 multiuser vs. frequency vs. spatial diversity in, 159–160
 random orthogonal, 211
 transmit selection diversity vs., 183
- bearer service architecture, 355–359
- bearers, QoS control, 355–359
- Bell Labs LAYered Space-Time (BLAST) receiver, 196–198
- BEP (bit-error probability)
 diversity gain and, 169–170
 modulation using SDNR, 123
 selection combining and, 172–173
- BER (bit error)
 in AMC, 87–88
 in OFDM vs. SC-FDE, 126
 in QAM modulation system, 80–81
 in unmitigated fading, 81–83
- bibliography
 cellular technologies, evolution of, 42
 downlink transport channels, 293–294
 frequency domain multiple access, 160–165
 inter-cell interference control, 381–382
 LTE radio interface, 255–256
 MIMO, 217–224
 mobility management, 381–382
 multicarrier modulation, 130–132
 physical layer procedures/scheduling, 353–354
 protocol architecture in LTE, 381–382
 uplink transport channels, 317
 wireless fundamentals, 93–97
- bit error. *See* BER (bit error)
- bit-error probability. *See* BEP (bit-error probability)
- bits
 scrambling codewords using, 263–264
 in sub-block interleaving, 263
 in wireless digital communication systems, 45–46
- BLAST (Bell Labs LAYered Space-Time) receiver, 196–198
- BLER (Block Error Rate), 82, 324
- Block Error Rate (BLER), 82, 324
- block transmission with guard levels, OFDM, 103
- BP (bandwidth part), UE-selected subband CQI report, 328–329
- broadband fading
 mitigation of, 90–92
 modelling channels, 67–80
 overview of, 60–67
- broadband wireless channels
 fading phenomenon. *See* fading
 overview, 46–48
 path loss, 48–51
 shadowing, 51–55
- Broadcast Channel. *See* BCH (Broadcast Channel)
- Broadcast Control Channel (BCCH), 236, 284

- BS (base stations)
 - in cellular systems, 55
 - early mobile telephone systems, 3
 - GSM subsystem, 8–9
 - LTE buffer status reporting and, 154
 - LTE channel models for path loss and, 76–77
 - networked MIMO coordinating, 212–213
 - OFDMA architecture, 141
 - BSC (base-station controller), evolution of, 8–9, 13
 - BSR (Buffer Status Report)
 - signaling with MAC control element, 368
 - in uplink scheduling, 154, 337–338
 - BSS (Base Station Subsystem), GSM, 8–9
 - BTS (base-station transceiver), 8–9, 13
 - Buffer Status Report. *See* BSR (Buffer Status Report)
- C**
- C Band, 38
 - C-RNTI, and semi-persistent scheduling, 345–346
 - cancellation, ICI, 377, 379
 - Carrier Sense Multiple Access (CSMA)
 - systems, 135–136, 142
 - carrier-to-interference ratio (CIR), 6
 - CAZAC (Constant Amplitude Zero Auto-Correlation) property, 289
 - CCCH (Common Control Channel) logical channel, 236
 - CCDF (complimentary cumulative distribution function), 54, 119–120
 - CCI (co-channel interference), 56–60
 - CDD (cyclic delay diversity)
 - large-delay, 273
 - open-loop spatial multiplexing, 214
 - CDMA (Code Division Multiple Access)
 - for 2G systems, 6–7
 - for 3G systems, 11–12
 - 2000-1X, 11–14
 - evolving networks to LTE, 16
 - HSPA vs. WiMAX vs. LTE and, 20
 - LTE co-existence with, 27
 - mitigating broadband fading, 90
 - in OFDM systems, 137–138
 - OFDM vs., 28–30
 - overview of, 134–135
 - CDMA (Code Division Multiple Access), IS-95
 - as 2G system, 6–7, 9–10
 - evolution beyond 3G, 16
 - IMT-MC standard for evolution to IX-EV-DO, 11
 - cell search process, physical layer, 346–348
 - cell-specific downlink reference signals, 285–287
 - cells, in cellular systems, 55–56
 - cellular systems
 - analysis of, 56–58
 - concept of, 55–56
 - development of, 3
 - overview of, 54–55
 - sectoring, 58–60
 - cellular technologies, evolution of
 - 3G broadband wireless, 10–15
 - 3GPP standards, 22–24
 - beyond 3G, 15–22
 - beyond LTE- future of mobile broadband, 39–41
 - case for LTE/SAE, 23–28
 - introduction, 1–2
 - LTE network architecture, 33–35
 - LTE spectrums and migration, 35–39
 - LTE technologies and features, 28–33
 - mobile broadband, 3–10
 - summary and conclusions, 41
 - Central Limit Theorem. *See* CLT (Central Limit Theorem)
 - CEPT (Conference of European Posts and Telegraphs), GSM, 7
 - CFI (Control Format Indicator)
 - in channel encoding and modulation, 281–282
 - in downlink control channels, 277, 280–282
 - in MAC layer, 238
 - CGF (Charging Gateway Function), GSM, 8
 - channel coding
 - code block concatenation, 263
 - code block segmentation, 259
 - CRC addition, 259
 - downlink transport channel
 - processing and, 257–258
 - HSPA vs. WiMAX vs. LTE, 21
 - overview of, 258–260
 - rate matching, 262–263
 - schemes for, 259–262
 - tail-biting convolutional coding, 260
 - for Uplink Control Information, 302–306
 - uplink transport channels, 296–297

- channel delay spread
 - coherence bandwidth and, 64
 - cyclic prefix and, 105–107
 - in ISI suppression, 100–101
 - severity of ISI and, 63
 - wireless channel parameters in, 47
- channel-dependent scheduling, 339
- channel encoding
 - downlink control channels, 279–282
 - downlink shared channels, 269–270
 - uplink shared channels, 299
- channel estimation
 - MIMO/MIMO-OFDM, 202–206
 - OFDM/SC-FDE, 127
- channel feedback, MIMO and
 - MIMO-OFDM, 206–208
- channel-independent scheduling, 339
- channel knowledge
 - linear interference suppression with complete, 189–192
 - linear interference suppression with statistical, 190–192
 - open-loop MIMO without, 194–198
- channel mapping
 - around downlink shared channel, 269–270
 - around uplink shared channel, 298
 - for downlink control channels, 277
 - LTE radio interface and, 240–241
 - for uplink control information, 302
- channel matrix, in spatial multiplexing, 193
- Channel Quality Indicator. *See* CQI (Channel Quality Indicator) feedback
- channel sounding, uplink, 337
- channel state information. *See* CSI (channel state information)
- channel state information at transmitter (CSIT), 206–208
- channel structure. *See* hierarchical channel structure, LTE
- Charging Gateway Function (CGF), GSM, 8
- chase combining, H-ARQ
 - defined, 14
 - procedures in MAC/RLC, 369
 - in type I H-ARQ, 86
- ciphering function, PDCP, 362
- CIR (carrier-to-interference ratio), 6
- circuit-switched data, original GSM, 8
- circular convolution
 - DFT and, 104–105
 - OFDM cyclic prefix creating, 106
- CL (closed-loop) MIMO
 - downlink shared channels, 273–275
 - overview of, 198–200
 - precoder for, 333–337
 - rank-1 precoding, 271, 274
 - spatial multiplexing. *See* spatial multiplexing, CL MIMO
 - uplink shared channels, 300
- CL (closed-loop) power control, LTE, 155
- CL (closed-loop) switched antenna diversity, 214–215
- CL (closed-loop) transmit diversity
 - defined, 174–175
 - with linear diversity precoding, 185–186
 - overview of, 181–182
 - with TSD, 182–185
- clean slate approach, LTE radio interface, 228–229
- clipping (soft limiting)
 - OFDM/SC-FDE design, 127
 - reducing PAR with, 121–123
- closed-loop (CL) MIMO. *See* CL (closed-loop) MIMO
- closed-loop (CL) power control, LTE, 155
- closed-loop (CL) switched antenna diversity, 214–215
- closed-loop transmit diversity. *See* CL (closed-loop) transmit diversity
- CLT (Central Limit Theorem)
 - lognormal distribution for shadowing and, 54
 - PAR and, 118
 - Rayleigh fading and, 68
- CN (core network), UMTS
 - in 3G evolution of GSM, 8
 - defined, 13
 - LTE radio interface and, 227
- co-channel interference (CCI), 56–60
- co-channel reuse ratio, 56–57
- code block concatenation, in channel coding, 263
- code block segmentation, in channel coding, 259
- Code Division Multiple Access. *See* CDMA (Code Division Multiple Access)
- codebooks
 - channel estimation and feedback for MIMO, 207
 - for CL MIMO modes, 335–336

- codebooks (*continued*)
 - in closed-loop spatial multiplexing, 214
 - defined, 333
 - in MU-MIMO, 214
 - precoding, 206
- codewords
 - in channel coding, 257–258
 - in MIMO transmission, 265–266
 - modulation mapping in downlink, 264
 - scrambling, 263–264
- coding and interleaving
 - exploiting diversity with, 201–202
 - mitigating narrowband fading, 83–85
 - OFDM vs. SC-FDE performance and, 126
- coding rate, 83
- coherence bandwidth
 - broadband fading parameters for, 63
 - delay spread and, 64
 - intersymbol interference approach to, 101–102
 - multicarriers and, 100
- coherence distance
 - angular spread and, 66–67
 - broadband fading parameters for, 63
- coherence times
 - broadband fading parameters for, 63
 - Doppler spread and, 65–66
- coherent demodulation, in OFDM, 29
- Common Control Channel (CCCH) logical channel, 236
- CoMP (Coordinated Multi-Point)
 - transmission/reception, LTE-Advanced, 378–379
- competitive market, and LTE
 - development, 25
- completion phase
 - S1 mobility, 373
 - X2 mobility, 374
- complimentary cumulative distribution function (CCDF), 54, 119–120
- computational complexity
 - of maximum likelihood detector, 194–195
 - OFDM/SC-FDE advantage, 124–125, 127–128
 - SC-FDMA/OFDMA disadvantage, 144
 - of transmit diversity, 28–29
- Conference of European Posts and Telegraphs (CEPT), GSM, 7
- connection control, RRC, 371
- Constant Amplitude Zero Auto-Correlation (CAZAC) property, 289
- contention resolution, random access channel, 350
- Control Format Indicator. *See* CFI (Control Format Indicator)
- control information, defined in MAC layer, 238–241
- Control PDU Type (CPT) field, STATUS PDU, 366
- control plane
 - latency in LTE design, 229
 - overview of, 356–358
 - PDCP functions, 360–361
 - PDCP integrity and ciphering in, 362
- convolutional coding
 - interleaving used with, 85
 - overview of, 83–84
- convolutional turbo encoding
 - channel coding in LTE, 260–262
 - channel coding scheme/coding rates, 259–260
 - downlink shared channels, 269
 - overview of, 260–262
- Coordinated Multi-Point (CoMP)
 - transmission/reception, LTE-Advanced, 380
- coordination, ICI, 378, 379
- cost efficiency, LTE design, 25, 27
- COST Hata model, 76–77
- COUNT parameter, PDCP security, 361–362
- coverage area
 - LTE design principles, 230
 - spatial diversity increasing, 171
- CP (cyclic prefix)
 - cell search process, 348
 - in downlink, 241
 - in LTE, 109–110
 - in OFDM, 109
 - overview of, 105–107
 - in SC-FDE, 125
 - in timing synchronization, 111–113
- CPT (Control PDU Type) field, STATUS PDU, 366
- CQI (Channel Quality Indicator) feedback
 - in aperiodic CQI reporting, 331–333
 - channel feedback and, 207
 - downlink scheduling/AMC-mode selection, 154
 - downlink shared channel modulation and, 269

- estimation process, 323–325
 - HSPDA cell phones and, 15
 - modes, overview, 325–326
 - MU-MIMO in uplink affecting, 301
 - overview of, 322
 - in periodic CQI reporting, 326–330
 - CQI/PMI, UCI on PUCCH, 302–305
 - CRC (cyclic redundancy check)
 - channel coding error detection using, 258–259
 - code block segmentation using, 259
 - DCI channel encoding and modulation using, 279–280
 - defined, 259
 - for PBCH error detection, 283
 - signaling, in downlink scheduling, 340–341
 - CSI (channel state information)
 - linear diversity precoding and, 185–186
 - MU-MIMO difficulties with, 275–276
 - obtaining at transmitter, 206–208
 - OFDMA architecture with, 141
 - OFDMA opportunistic scheduling and, 146–147
 - SDMA vs. SU-MIMO sensitivity to, 212
 - CSIT (channel state information at transmitter), 206–208
 - CSMA (Carrier Sense Multiple Access) systems, 135–136, 142
 - cyclic delay diversity (CDD), 214, 273
 - cyclic prefix. *See* CP (cyclic prefix)
 - cyclic redundancy check. *See* CRC (cyclic redundancy check)
- D**
- D-BLAST (diagonal BLAST), open-loop MIMO, 196–197
 - D/C (Data/Control) field
 - PDCP data PDU for user plane, 360–361
 - RLC PDUs, 366
 - DAC (digital-to-analog) converters, 46, 116–117
 - Data Radio Bearers (DRBs), 355
 - data rate
 - 2G limitations, 7
 - 3G providing higher, 10
 - 3GPP standards evolution, 22–24
 - advantages of OFDMA lower, 142
 - CDMA2000-1X as 2.5 system, 12
 - CDMA2000-1X-EVDO improving, 12–13
 - GPRS moving towards higher, 8
 - HSPA+ enhancements to, 17
 - HSPA vs. WiMAX vs. LTE, 20–22
 - linear diversity precoding increasing, 185–186
 - LTE-Advanced requirements, 40
 - LTE design principles, 26, 229
 - spatial diversity increasing, 170–171
 - data transfer modes, RLC layer, 363–364
 - DCCH (Dedicated Control Channel)
 - logical channel, 236
 - DCF (Distributed Coordination Function), CSMA, 135
 - DCI (Downlink Control Information)
 - channel encoding and modulation, 279–280
 - defined in MAC layer, 238
 - downlink control channels, 276–279
 - downlink shared channels, 270
 - formats, 277–279
 - semi-persistent scheduling for VoIP, 345–346
 - signaling for downlink scheduling, 340–341
 - deciphering function, PDCP, 362
 - decision feedback equalizers (DFE), 91
 - DECT (Digital Enhanced Cordless Telephone), 6
 - dedicated bearer, 357
 - Dedicated Control Channel (DCCH)
 - logical channel, 236
 - Dedicated Traffic Channel (DTCH) logical channel, 236
 - default bearer, 357
 - delay dimension, broadband fading channels, 62–63
 - delay spread
 - broadband fading parameters for, 63
 - coherence bandwidth and, 64
 - OFDM robustness against, 29
 - delay spread (DS), multipath modelling, 78
 - demodulation reference signals, 309–312
 - deployment, LTE requirements, 230
 - DFE (decision feedback equalizers), 91
 - DFT (Discrete Fourier Transform), OFDM, 103–105, 297–298
 - diagonal BLAST (D-BLAST), open-loop MIMO, 196–197
 - Digital Enhanced Cordless Telephone (DECT), 6

- digital modulation, 2G systems, 6
- Digital Subscriber Line (DSL), and multicarrier modulation, 99
- digital-to-analog (DAC) converters, 46, 116–117
- Digital Video Broadcasting, 99
- Direct Sequence Spread Spectrum (DSSS), W-CDMA as, 13–14
- direction of arrival (DOA)-based beamforming, 187–189, 251
- directional antennas, sectoring with, 59
- Discontinuous Reception (DRX), Paging function of, 377
- Discrete Fourier Transform (DFT), OFDM, 103–105, 297–298
- Discrete Multitone (DMT) in DSL, 99
- discrete-time channel, broadband wireless model, 47–48
- dispersion-selectivity duality, 74–75
- Distributed Coordination Function (DCF), CSMA, 135
- distributed subcarrier allocation, 153
- distributed VRBs (virtual resource blocks), 250–251
- diversity
 - combining narrowband techniques for, 89
 - defined, 80
 - frequency. *See* frequency diversity
 - interference suppression vs. spatial multiplexing vs., 200–202
 - multiuser vs. frequency vs. spatial diversity, 159–160
 - OFDMA multiuser, 144–146
 - receive diversity, 171–174
 - spatial. *See* spatial diversity
 - transmit. *See* transmit diversity
 - unmitigated fading affecting, 80–83
- diversity gain
 - defined, 168
 - Diversity-Multiplexing Tradeoff, 200–201
 - multiantennas and, 169–170
- Diversity-Multiplexing Tradeoff (DMT), 200–201
- DL-SCHs (Downlink Shared Channels)
 - channel encoding and modulation, 269–270
 - CL MIMO techniques, 273–275
 - downlink transport channel processing, 268–276
 - MU-MIMO in downlink, 275–276
 - multiantenna transmission, 270–271
 - OL MIMO techniques, 271–273
 - overview of, 268–269
 - random access response on, 349
 - as transport channel, 237
- DMT (Discrete Multitone) in DSL, 99
- DMT (Diversity-Multiplexing Tradeoff), 200–201
- DOA (direction of arrival)-based beamforming, 187–189, 251
- Doppler spread
 - broadband fading parameters for, 63
 - and coherence time, 65–66
 - time correlation of received signal, 71–72
- double space-time transmit diversity (DSTTD), stacked STBCs, 178–179
- downlink
 - allocation notification in, 154
 - CDMA2000-1X improvements to, 12
 - CQI and UCI in, 301
 - CQI feedback mechanism in, 322
 - designated FDD frequency bands for, 35–36
 - H-ARQ transmissions in, 320–321
 - High-Speed Packet Access in, 14–15
 - HSPA+ enhancements to, 16–17
 - HSPA vs. WiMAX vs. LTE in, 20–22
 - ICI coordination in LTE, 377–379
 - Layer 2 structure in, 358
 - LTE-Advanced requirements, 40
 - LTE design principles, 26, 229
 - MIMO in LTE, 213–214
 - OFDM for core modulation in, ?
 - OFDMA in LTE, 139–141, 152–155
 - physical channels, 239–240
 - power control in LTE, 154–155
 - SDMA vs. SU-MIMO, 211
 - signaling for scheduling in, 340–341
- downlink control channels
 - Control Format Indicator, 280–282
 - DCI channel encoding and modulation, 279–280
 - DCI formats, 277–279
 - H-ARQ Indicator, 282
 - multiantenna transmission, 282–283
 - overview of, 276–277
- Downlink Control Information. *See* DCI (Downlink Control Information)
- downlink OFDMA radio resources
 - frame structure, 242–243
 - frame structure type 1, 243–244

frame structure type 2, 244–246
 overview of, 241–242
 physical resource blocks for OFDMA,
 246–248
 resource allocation, 248–251
 supported MIMO modes, 251
 downlink physical signals
 cell-specific reference signals,
 285–287
 downlink reference signals, 285
 MBSFN reference signals, 287
 synchronization signals, 289–290
 UE-specific reference signals, 287–288
 Downlink Pilot TimeSlot (DwPTS) field,
 245, 252
 downlink reference signals
 cell-specific, 285–287
 MBSFN, 287
 overview of, 285
 UE-specific, 287–288
 Downlink Shared Channels. *See* DL-SCHs
 (Downlink Shared Channels)
 downlink transport channel processing
 broadcast channels, 283
 channel coding. *See* channel coding
 downlink control channels, 276–283
 downlink physical signals, 285–290
 downlink shared channels. *See*
 DL-SCHs (Downlink Shared
 Channels)
 H-ARQ in downlink, 290–292
 modulation, 263–268
 multicast channels, 284–285
 overview of, 257–258
 summary and conclusions, 293
 downlink transport channels, 237
 DRBs (Data Radio Bearers), 355
 DRX (Discontinuous Reception), Paging
 function of, 377
 DS (delay spread), multipath modelling, 78
 DSL (Digital Subscriber Line), and
 multicarrier modulation, 99
 DSSS (Direct Sequence Spread Spectrum),
 W-CDMA as, 13–14
 DSTTD (double space-time transmit
 diversity), stacked STBCs,
 178–179
 DTCH (Dedicated Traffic Channel) logical
 channel, 236
 dual-carrier operation, HSPA+ and, 17
 dual layer beamforming, 276
 DwPTS (Downlink Pilot TimeSlot) field,
 245, 252

E

E-DCH (Enhanced Dedicated Channel),
 HSUPA, 15
 E (Extension bit) field, MAC subheader,
 367
 E (Extension bit) field, RLC headers, 365
 E-MBMS (Enhanced MBMS), 284
 E-UTRA (Evolved UMTS Terrestrial
 Radio Access), 227
 E-UTRAN (Enhanced UTRAN)
 development of, 24
 LTE network architecture, 231–232
 network architecture, 229
 overview of, 227
 Paging function, 376–377
 RAN procedures for mobility,
 375–376
 E1 (Extension bit 1) field, STATUS PDU,
 366
 E2 (Extension bit 2) field, STATUS PDU,
 366
 Earthquake and Tsunami Warning System
 (ETWS), 376
 ECCs (error correction codes), 83–85
 EDGE (Enhanced Data Rate for GSM
 Evolution), 9, 22
 effective channel coding rate, downlink
 shared channels, 269
 effective SINR, in CQI estimation,
 322–324
 EGC (equal gain combining), 174
 empirical channel models
 defined, 67
 functionality of, 75–76
 LTE models for multipath, 77–79
 LTE models for path loss, 76–77
 LTE semi-empirical models, 79–80
 empirical path loss formula, 50–52
 encoding, wireless digital systems, 45–46
 encryption, in 2G systems, 6
 Enhanced Data Rate for GSM Evolution
 (EDGE), 9, 22
 Enhanced Dedicated Channel (E-DCH),
 HSUPA, 15
 Enhanced MBMS (E-MBMS), 284
 Enhanced Uplink, 15
 eNode-B
 CQI estimation process, 324
 defined, 32
 LTE flat network architecture and,
 227–228, 231
 MAC entities in, 364
 MU-MIMO in uplink affecting, 301

- eNode-B (*continued*)
 - non-synchronized random access procedure and, 348–350
 - overriding PMI reported by UEs, 336–337
 - resource allocation in uplink at, 254
 - S1 mobility, 371–373
 - semi-persistent scheduling for VoIP, 344–346
 - signaling for downlink scheduling, 340–341
 - signaling for uplink scheduling, 341–343
 - uplink channel sounding at, 337
 - X2 mobility at, 232, 373–374
 - EPC (Evolved Packet Core)
 - architecture, 33–34
 - development of, 24
 - LTE network architecture
 - components, 231–232
 - LTE project in 3GPP enhancing, 227
 - protocol architecture in LTE, 356–357
 - services in LTE flat architecture, 32
 - EPS (Evolved Packet System)
 - bearer, 355–359, 380
 - defined, 227
 - development of, 24
 - equal gain combining (EGC), 174
 - equalizers
 - ISI-suppression to OFDM and, 91
 - OFDM/SC-FDE advantages, 127–128
 - error correction codes (ECCs), 83–85
 - errors
 - detecting for broadcast channels, 283
 - H-ARQ handling transmission, 369
 - nonlinear equalizers causing, 91
 - estimation process, CQI, 322–325
 - ETACS (Total Access Communication System), 5
 - Ethernet, CSMA for random access, 135–136
 - ETSI (European Telecommunications Standards Institute), 7–8
 - ETWS (Earthquake and Tsunami Warning System) notification, 376
 - European Telecommunications Standards Institute (ETSI), 7–8
 - EV-DO, 11–13, 229
 - events, intra-LTE and intra-RAT handover, 375
 - Evolved Packet Core. *See* EPC (Evolved Packet Core)
 - Evolved Packet System. *See* EPS (Evolved Packet System)
 - Evolved UMTS Terrestrial Radio Access (E-UTRA), 227
 - execution phase
 - S1 mobility, 372
 - X2 mobility, 374
 - Extended delay profiles, 79–80
 - Extended Pedestrian A delay profile, 79
 - Extended Vehicular A delay profile, 79
 - Extension bit 1 (E1) field, STATUS PDU, 366
 - Extension bit 2 (E2) field, STATUS PDU, 366
 - Extension bit (E) field, MAC subheader, 367
 - Extension bit (E) field, RLC headers, 365
- ## F
- F field, MAC subheader, 367
 - fading
 - empirical/semi-empirical models
 - estimating, 76
 - large-scale. *See* large-scale fading mitigation of narrowband. *See* narrowband fading, mitigation of
 - fading, broadband
 - angular spread/coherence distance in, 66–67
 - delay spread/coherence bandwidth in, 64
 - Doppler spread/coherence time in, 65–66
 - empirical channel models for, 75–80
 - mitigation of, 90–92
 - overview of, 60–63
 - statistical channel models for, 68–71
 - statistical correlation of received signal for, 71–75
 - fairness
 - Maximum Fairness algorithm, 148
 - in OFDMA resource allocation, 150–152
 - power control in, 155
 - fast dynamic scheduling, HSPDA/HSUPA, 14
 - Fast Fourier Transforms. *See* FFT (Fast Fourier Transforms)
 - FCC (Federal Communications Commission), cellular mobile concept, 3

- FDD (Frequency Division Duplexing)
 - mode
 - aperiodic CQI reporting in, 331–332
 - designated frequency bands for LTE in, 35–36
 - frame structure type 1 in, 243–244
 - H-ARQ in downlink, 320
 - H-ARQ in uplink, 315, 321
 - LTE support for, 27, 230
 - resource mapping of sounding reference signals, 311–313
 - UCI on PUSCH with UL-SCH data in, 304–305
 - W-CDMA for, 13–14
 - WiMAX for, 19
- FDMA (Frequency Division Multiple Access)
 - CSMA vs., 136
 - implementing in OFDM systems, 136–137
 - overview of, 134–135
 - single-carrier. *See* SC-FDMA (Single Carrier Frequency Division Multiple Access)
- FEC (forward error correction) code
 - channel coding error-control, 258
 - at WiMAX link layer, 19
- Federal Communications Commission (FCC), cellular mobile concept, 3
- FEQ (frequency domain equalizer), OFDM
 - overview of, 108
 - SC-FDE vs., 125–126
 - steps in communications, 109
- FER (Frame Error Rate), and unmitigated fading, 82
- FFR (Fractional Frequency Reuse), mitigating ICI, 157–159
- FFT (Fast Fourier Transforms)
 - cyclic prefix and, 105
 - frame structure in time domain, 242
 - implementing OFDM communications with, 108
 - OFDM baseband signal generated with, 268
 - OFDM implementation with, 28, 103
 - OFDM vs. SC-FDE, 125
 - SC-FDMA and, 91
 - WiMAX scalability and, 19
- FFT (IFast Fourier Transform)
 - OFDM/SC-FDE computational complexity advantage, 128
 - OFDM/SC-FDE design, 127
- FH (Frequency Hopping) field, PUSCH, 299–300
- FI (Framing Info) field, RLC headers, 365
- fixed error probability, DMT theory, 201
- fixed line subscribers, 1
- flat fading
 - converting frequency-selective fading in MIMO to, 208
 - defined, 74
 - intersymbol interference approach, 101–103
 - multicarriers and, 100
- flat network architecture, LTE
 - 3GPP evolution towards, 32–33
 - overview of, 227–228
- flat-rate pricing, in LTE development, 25
- forward error correction (FEC) code
 - channel coding error-control, 258
 - at WiMAX link layer, 19
- fourth generation systems. *See* 4G (fourth generation) systems
- FPC (Fractional Power Control), in uplink, 351
- Fractional Frequency Reuse (FFR), mitigating ICI, 157–159
- Fractional Frequency Reuse (Strict FFR), mitigating ICI, 157–158
- Fractional Power Control (FPC), in uplink, 351
- fractional power control, LTE, 155
- Frame Error Rate (FER), and unmitigated fading, 82
- frame size, HSPA vs. WiMAX vs. LTE, 21–22
- frame structure
 - downlink OFDMA radio resources, 242–246
 - mapping synchronization signals, 289–290
 - random access preambles, 314
 - uplink SC-FDMA, 252
- Framing Info (FI) field, RLC headers, 365
- free-space path loss (Friss) formula, 48–49
- frequency correlation, of received signal, 72–74
- frequency, dispersion-selectivity in, 74
- frequency diverse scheduling, 339
- frequency diversity
 - for maximal ratio combining, 174
 - OFDMA advantages, 29, 142
 - spatial diversity vs. multiuser diversity vs., 159–160
- Frequency Division Duplexing. *See* FDD (Frequency Division Duplexing) mode

- Frequency Division Multiple Access. *See* FDMA (Frequency Division Multiple Access)
- frequency domain
- cell-specific reference signals in, 287
 - CQI reporting in, 325
 - generating random access in, 314
 - MBSFN reference signals in, 287
 - resource allocation in, 249
- frequency-domain channel estimation, MIMO-OFDM, 205–206
- frequency domain equalizer. *See* FEQ (frequency domain equalizer), OFDM
- frequency domain multiple access
- Maximum Fairness algorithm, 148
 - Maximum Sum Rate algorithm, 147–148
 - multiple access for OFDM systems, 134–138
 - multiuser diversity and, 144–146
 - OFDMA, 138–141
 - OFDMA and SC-FDMA in LTE, 152–155
 - OFDMA design considerations, 156–160
 - OFDMA opportunistic scheduling, 146–147
 - overview of, 133–134
 - performance comparison, 150–152
 - Proportional Fairness scheduling, 149–150
 - Proportional Rate Constraints algorithm, 148–149
 - SC-FDMA, 142–144
 - summary, 160
- frequency equalization, 107–109, 127–128
- frequency hopping, 90, 299–300
- Frequency Hopping (FH) field, PUSCH, 299–300
- Frequency Modulation, AMPS, 6
- frequency planning
- cellular performance and, 58
 - for cellular systems, 55–56
 - defined, 55
- frequency reuse
- intelligent planning of, 55–56
 - mitigating ICI with, 157
 - sacrificing bandwidth by using, 58
 - using sectoring with, 59
- frequency selection, 20, 26–27
- frequency-selective fading
- converting to flat-fading with OFDM, 208
 - defined, 74
 - DMT theory for, 201
- frequency selective scheduling, 339–341
- Frequency Shift Keying (FSK), AMPS, 6
- frequency shift time diversity (FSTD), 31, 179–180
- frequency synchronization, 110–112, 114–116
- Friss (free-space path loss) formula, 48–49
- FSK (Frequency Shift Keying), AMPS, 6
- FSTD (frequency shift time diversity), 31, 179–180
- ## G
- gains
- combining narrowband diversity techniques for greater, 89
 - issues limiting MIMO, 208
- Gateway GPRS Support Node (GGSN), 8–9, 32
- Gaussian Minimum Shift Keying (GMSK), 8
- GBR (Guaranteed Bit Rate) bearers, 356
- GERAN (GSM EDGE radio access networks), and LTE, 230
- GGSN (Gateway GPRS Support Node), 8–9, 32
- Global System for Mobile Communications. *See* GSM (Global System for Mobile Communications)
- GMSK (Gaussian Minimum Shift Keying), 8
- GP (Guard Period) field
- downlink OFDMA radio resources, 245
 - uplink SC-FDMA radio resources, 252
- GPRS (GSM Packet Radio Systems)
- latency in, 22
 - UMTS compatibility with, 13
 - upgrading GSM system to, 8–9
- GPRS Tunneling Protocol (GTP), 357
- Groupe Special Mobile, 7–8
- GSM EDGE radio access networks (GERAN), and LTE, 230
- GSM (Global System for Mobile Communications)
- as 2G system, 6–7
 - designated frequency band for LTE, 35–37

- evolution of, 7–9
 - UMTS compatibility with, 13
- GSM Packet Radio Systems. *See* GPRS (GSM Packet Radio Systems)
- GTP (GPRS Tunneling Protocol), 357
- Guaranteed Bit Rate (GBR) bearers, 356
- guard levels, OFDM
 - block transmission with, 103
 - and cyclic prefix, 105
 - OFDM in LTE, 110
- Guard Period (GP) field
 - downlink OFDMA radio resources, 245
 - uplink SC-FDMA radio resources, 252
- H**
- H-ARQ-ACK
 - feedback for downlink transmission, 320–321
 - modulation of PUCCH, 308
 - procedures in MAC/RLC, 369
 - UCI on PUCCH, 303–304
 - UCI on PUSCH with UL-SCH data, 304–306
- H-ARQ (Hybrid-ARQ) protocol
 - channel coding error-control using, 259
 - CQI estimation process, 324
 - in downlink, 290–292, 320–321
 - in Downlink Shared Channel, 237
 - HSPA supporting, 15
 - HSPA vs. WiMAX vs. LTE, 21
 - HSPDA/HSUPA supporting, 14–15
 - MAC protocol error correction through, 234
 - MAC/RLC procedures, 368–369
 - mitigating narrowband fading, 86
 - SAPs between different layers, 234–235
 - for semi-persistent scheduling in VoIP, 345–346
 - Uplink Control Information and, 301
 - WiMAX link layer supporting, 18
- H-ARQ Indicator. *See* HI (H-ARQ Indicator)
- handover performance
 - in IS-95 CDMA systems, 9
 - RAN procedures for mobility, 374–376
 - Si mobility, 372–373
 - X2 mobility, 373–374
- Hata Model for Suburban and Open Areas, 77
- Hata Model for Urban Areas, 77
- HDR (High-Data Rate), EV-DO
 - evolution, 13
- header compression, PDCP, 361
- headers, RLC PDU formats, 365–366
- hexagonal cell structure, 55–57
- HFN (Hyper Frame Number), PDCP, 361–362
- HI (H-ARQ Indicator)
 - channel encoding and modulation, 282
 - defined in MAC layer, 238
 - downlink control channels containing, 277, 282
 - for uplink transmission, 321
- hierarchical cell search scheme, LTE, 346–348
- hierarchical channel structure, LTE
 - channel mapping, 240–241
 - logical channels, 235–236
 - overview of, 234–235
 - physical channels, 239
 - transport channels, 236–237
 - uplink physical channels, 239–240
 - uplink transport channels, 237–239
- High-Data Rate (HDR), EV-DO
 - evolution, 13
- High Interference Indicator (HII) messages, mitigating uplink ICI, 380
- high-power amplifier (HPA), PAR problem and, 116–117
- High-Speed Downlink Packet Access (HSDPA), 14–15, 229–230
- High-Speed Downlink Packet Access (HSPDA), 228, 237
- High-Speed Downlink Shared Channel (HS-DSCH), 14, 237
- High-Speed Packet Access. *See* HSPA (High-Speed Packet Access)+
- High-Speed Uplink Packet Access. *See* HSUPA (High-Speed Uplink Packet Access)
- higher layer configured subband CQI report
 - aperiodic CQI reporting, 325, 331–332
 - PMI feedback type, 327
- HII (High Interference Indicator) messages, mitigating uplink ICI, 380
- HLR (Home Location Register), GSM, 8–9
- Home Location Register (HLR), GSM, 8–9
- Home Subscriber Service (HSS), 232

- HPA (high-power amplifier), PAR problem and, 116–117
- HS-DSCH (High-Speed Downlink Shared Channel), 14, 237
- HSDPA (High-Speed Downlink Packet Access), 14–15, 229–230
- HSPA (High-Speed Packet Access)+
 evolution beyond 3G, 15
 H-ARQ of MAC/RLC layers in, 369
 LTE radio interface design vs., 228
 overview of, 14–15
 packet-switched network support in, 229
 reduced latency of, 22
 technical enhancements of, 16–17
 WiMAX vs. LTE vs., 20–22
- HSPDA (High-Speed Downlink Packet Access), 228, 237
- HSS (Home Subscriber Service), 232
- HSUPA (High-Speed Uplink Packet Access)
 defined, 14
 HSPA as collection of, 228
 LTE spectrum efficiency vs., 229–230
 overview of, 15
- Hybrid-ARQ. *See* H-ARQ (Hybrid-ARQ) protocol
- Hyper Frame Number (HFN), PDCP, 361–362
- ## I
- IBO (input backoff), 116–117, 121–123
- ICI (Inter-Cell Interference) coordination in LTE
 in downlink, 377–379
 mitigation techniques, 157–159
 OFDMA/SC-FDMA and, 154
 overview of, 377
 summary and conclusions, 381
 in uplink, 379–380
- IDFT (Inverse Discrete Fourier Transform), 104–107
- IEEE 802.16e. *See* Mobile WiMAX
- IFFT (Inverse Fast Fourier Transform)
 cyclic prefix and, 105
 defining PAR for output of, 118–120
 OFDM baseband signal generation, 268
 OFDM implementation with, 28, 103, 108
 OFDM in LTE, 109–110
- OFDM/SC-FDE computational complexity advantage, 128
- OFDM/SC-FDE design, 127
- OFDM vs. SC-FDE, 125–126
- SC-FDE in LTE and, 30
- SC-FDMA and, 91
 time correlation of received signal, 72
- improved Mobile Telephone System (IMTS), 3
- IMS (IP Multimedia Subsystem) domain, 232
- IMT-2000 CDMA Direct Spread (IMT-DS) standard. *See also* W-CDMA (wideband CDMA), 11
- IMT-2000 CDMA Multi-carrier (IMT-MC) standard, 11
- IMT-2000 CDMA TDD (IMT-TC) standard, 3G, 11
- IMT-2000 FDMA/TDMA (IMT-FT) standard, 11
- IMT-2000 IP-OFDMA standard. *See also* WiMAX, 11
- IMT-2000 TDMA Single Carrier (IMT-SC) standard, 3G, 11
- IMT-Advanced, 4G, 40
- IMT-DS (IMT-2000 CDMA Direct Spread) standard. *See also* W-CDMA (wideband CDMA), 11
- IMT extension band, early LTE, 39–40
- IMT-FT (IMT-2000 FDMA/TDMA) standard, 11
- IMT-MC (IMT-2000 CDMA Multi-carrier) standard, 11
- IMT-SC (IMT-2000 TDMA Single Carrier) standard, 3G, 11
- IMT-TC (IMT-2000 CDMA TDD) standard, 3G, 11
- IMTS (improved Mobile Telephone System), 3
- in-band distortion, by reducing PAR with clipping, 122
- incremental redundancy, H-ARQ
 defined, 14
 MAC/RLC procedures, 369
 in type II H-ARQ, 86
- input backoff (IBO), 116–117, 121–123
- integrity protection, PDCP, 361–362
- Inter-Cell Interference. *See* ICI (Inter-Cell Interference) coordination in LTE
- inter-subframe frequency hopping, LTE, 299
- interference-aware beamforming, 191

- interference channels, knowledge of, 189–192
- interference-limited systems, 56–57, 209
- interference suppression
 - diversity vs. spatial multiplexing vs., 200–202
 - DOA-based beamsteering for, 187–189
 - linear, 189–192
 - multiantennas and, 186–187
- interleaving, and rate matching, 262–263
- International Telecommunications Union (ITU), 10–11, 18
- Internet access
 - global growth of broadband for, 1–2
 - limitations of 2G, 10
 - LTE development and, 24–25
 - LTE requirements, 26–27
- interoperability with 3G and 2G, LTE, 231
- intersector handoffs, mitigating OCI, 59–60
- intersymbol interference. *See* ISI (intersymbol interference) mitigation
- intra-LTE handover events, 374–376
- intra-LTE mobility, 371–374
- intra-RAT handover events, 375
- intra-subframe frequency hopping, LTE, 299
- Inverse Discrete Fourier Transform (IDFT), 104–107
- Inverse Fast Fourier Transform. *See* IFFT (Inverse Fast Fourier Transform)
- IP Multimedia Subsystem domain (IMS), 232
- IP protocols
 - LTE flat network based on, 32–33
 - LTE/SAE and, 23–24
 - for services in LTE flat architecture, 32
 - WiMAX reliance on, 20
- ISI (intersymbol interference) mitigation
 - circular convolution and DFT, 104–105
 - elegant approach to, 101–103
 - equalizers for, 91
 - multicarrier concept and, 99–101
 - OFDM block transmission with guard levels for, 103
 - OFDM communication steps and, 108
 - OFDM multicarrier modulation for, 91, 99
 - overview of, 90
 - SC-FDMA for, 91
 - spread spectrum and RAKE receivers for, 90
- ITU (International Telecommunications Union), 10–11, 18
- IX-EV-DO, 11–13
- L**
- L field, MAC subheader, 367–368
- large-scale fading, 52
- Last Segment Flag (LSF) field, RLC PDU, 366
- latency
 - 3GPP standards evolution, 22–24
 - LTE-Advanced requirements, 40
 - LTE design requirement of low, 26–27, 229
- layer 1 of OSI reference model, PHY (physical layer) as, 234
- Layer 2 of LTE protocol stack, 357–358, 380
- layer mapping, 264–266
- layers, in MIMO transmission, 265–266
- LCID (Logical Channel ID) field, MAC subheader, 367
- least-squares channel estimation, 205
- legacy systems, frame structure for, 244–246
- Length Indicator (LI) field, RLC headers, 365
- line-of-sight path. *See* LOS (line-of-sight) path
- linear detectors, OL MIMO, 195–196
- linear diversity precoding, 185–186
- linear equalizers, suppressing ISI, 91
- linear interference suppression, 189–192
- linear precoding
 - and postcoding, 198–200
 - with quantized feedback, 206–207
- links
 - WiMAX robust link layer, 19
 - wireless digital communication systems, 45–46
- localized VRBs (virtual resource blocks), 250–251
- logical CCCH (Common Control Channel), 236
- Logical Channel ID (LCID) field, MAC subheader, 367
- logical channels
 - defined, 234–235
 - logical control channels, 236

- logical channels (*continued*)
 - logical traffic channels, 236
 - mapping between transport channels and, 240–241
 - what to transmit, 235–236
 - lognormal distribution for shadowing, 54
 - Long-Term Evolution. *See* LTE (Long-Term Evolution)
 - LOS (line-of-sight) channels, 69–70
 - LOS (line-of-sight) path
 - causing temporary degradation in signal strength, 51
 - modeling Ricean distribution for, 69–70
 - using shadowing effect for, 51–55
 - LSF (Last Segment Flag) field, RLC PDUs, 366
 - LTE Advanced, 40–41
 - LTE-Advanced, 378–379
 - LTE (Long-Term Evolution)
 - as 4G system, 16
 - design requirements, 26–27
 - downlink transport channel
 - processing. *See* downlink transport channel processing
 - evolving networks to, 16
 - future enhancements, 39–41
 - for high-speed data services, 228–229
 - HSPA+ vs. WiMAX vs., 20–22
 - mobile broadband and, 2
 - multicarrier modulation
 - underlying, 99
 - network architecture of, 33–35
 - OFDM in, 109–110
 - OFDMA and SC-FDMA in, 152–155
 - open-loop transmit diversity in, 179–180
 - reduced latency of, 22
 - spectrum options and migration plans, 35–39
 - technologies/features of, 28–32
 - WiMAX inspiring design for, 18
 - LTE (Long-Term Evolution) radio interface
 - channel mapping, 240–241
 - design principles, 229–231
 - downlink OFDMA. *See* downlink OFDMA radio resources
 - hierarchical channel structure of, 234–240
 - introduction to, 228–229
 - network architecture, 231–232
 - overview of, 227–228
 - protocols, 232–234
 - summary and conclusions, 255
 - uplink SC-FDMA radio resources, 251–255
 - LTE/SAE, 23–28
- ## M
- M* subpaths, multipath modelling, 77–79
 - MAC-I field, integrity protection for PDCP, 362
 - MAC (Medium Access Control) layer
 - channel mapping at, 240–241
 - control information defined in, 238
 - HSPA+ enhancements to, 17–18
 - interaction with RLC. *See* MAC/RLC overview
 - Layer 2 of LTE protocol stack, 357–358
 - logical channels used by, 235–236
 - SAPs between different layers, 234–235
 - transport channel services for, 236–237
 - MAC RAR (random access response), 368
 - MAC/RLC overview
 - ARQ procedures, 368–369
 - data transfer modes, 363–364
 - MAC PDU formats, 367–368
 - PDU headers and formats, 365–367
 - purpose of, 364–365
 - mapping
 - channel. *See* channel mapping
 - channel coding and, 257–258
 - modulation in downlink, 264
 - modulation in uplink, 297–298
 - precoding and layer, 264
 - resource. *See* resource mapping
 - Marconi, Guglielmo, 3
 - Master Information Block. *See* MIB (Master Information Block)
 - matrix theory, 184–185
 - Max-Min problem, 148
 - maximal radio transmission (MRT), 191–192
 - maximal ratio combining. *See* MRC (maximal ratio combining)
 - maximum Doppler, 65
 - Maximum Fairness (MF) algorithm, 148, 150–152
 - maximum likelihood (ML) detector, OL MIMO, 194–195
 - Maximum Likelihood Sequence Detection (MLSD), suppressing ISI, 91

- Maximum Sum Rate (MSR) algorithm, 147–148, 150–152
- MB, LTE design reducing cost in, 25, 27
- MBMS (Multimedia Broadcast and Multicast Service)
 - downlink transport channel processing, 284–285
 - HSPA+ enhancements to, 18
 - LTE design principles, 230
- MBSFN (Multicast/Broadcast Single Frequency Network)
 - downlink reference signals, 287
 - frame structure, 242
 - Multicast Channel supporting, 237
 - overview of, 284–286
- MC-CDMA (Multicarrier CDMA), in OFDM systems, 137–138
- MCCH (Multicast Control Channel) logical channel, 236–237
- MCH (Multicast Channel) downlink channel, 237, 284–285
- MCS (Modulation and Coding Scheme), CQI, 322, 324
- Medium Access Control. *See* MAC (Medium Access Control) layer
- MF (Maximum Fairness) algorithm, 148, 150–152
- MIB (Master Information Block)
 - broadcast channels, 283
 - cell search process, 348
 - functions of RRC protocol, 371
- migration, options for LTE, 35–39
- milestones, mobile broadband, 4
- MIMO (multiantenna transmission and reception), 167–224
 - 3GPP channel model simulating, 78
 - angular spread/coherence distance in LTE, 66–67
 - channel estimation, 202–206
 - channel feedback, 206–208
 - closed-looped, 198–200
 - diversity vs. interference suppression vs. spatial multiplexing, 200–202
 - downlink control channels with, 282
 - downlink shared channels with, 270–271
 - HSPA+ enhancements to, 16
 - HSPA vs. WiMAX vs. LTE, 21–22
 - interference cancellation
 - suppression/signal enhancement in, 186–192
 - layer mapping and precoding in, 264–266
 - limiting gains, 208–209
 - in LTE, 31–32, 213–215, 228
 - in LTE downlink, 213–214, 251
 - in LTE uplink, 214–215, 254–255
 - multiuser. *See* MU-MIMO (multiuser MIMO)
 - multiuser diversity vs. frequency diversity vs. spatial diversity in, 159–160
 - networked systems, 212–213
 - OFDM combined with, 29–30
 - open-loop. *See* OL (open-loop) MIMO
 - overview of, 167–168
 - for PBCH, 283
 - receive diversity and, 171–174
 - spatial diversity and, 168–171
 - spatial multiplexing and. *See* spatial multiplexing
 - summary and conclusions, 215–216
 - supported modes for different physical channels, 266–267
 - transmit diversity, 174–186
- MIMO-OFDM
 - channel estimation for, 202–206
 - channel feedback for, 206–208
 - converting frequency-selective fading to flat-fading in, 208
- Minimum Mean Square Error receivers. *See* MMSE (Minimum Mean Square Error) receivers
- ML (maximum likelihood) detector, OL MIMO, 194–195
- MLSD (Maximum Likelihood Sequence Detection), suppressing ISI, 91
- MME (Mobility Management Entity)
 - control plane protocol stack in, 357
 - defined, 32
 - EPC architecture and, 33–35
 - LTE network architecture and, 231
 - Si mobility, 371–373
 - X2 mobility, 373–374
- MMSE (Minimum Mean Square Error) receivers
 - channel estimation for, 205–206
 - open-loop MIMO for, 196
 - precoder estimation for, 334
- mobile broadband
 - 2G systems, 6–10
 - first generation systems, 5
 - future of, 39–41
 - historical milestones in, 4
 - overview of, 3–4

- mobile stations. *See* MS (mobile stations)
- mobile subscribers, global growth of, 1–2
- Mobile Switching Center (MSC), GSM, 8–9
- Mobile Telephone System (MTS), 3
- Mobile WiMAX, 18–20
- Mobility Management Entity. *See* MME (Mobility Management Entity)
- mobility management in LTE
 - LTE-Advanced requirements, 40
 - overview of, 371
 - paging, 376–377
 - RAN procedures for, 374–376
 - requirement for, 230
 - S1 mobility, 371–373
 - summary and conclusions, 380–381
 - X2 mobility, 373–374
- modulation. *See also* multicarrier modulation
 - downlink control channels, 279–282
 - downlink shared channels, 269–270
 - HSPA vs. WiMAX vs. LTE, 20–21
 - of PUCCH, 306–308
 - signaling for downlink scheduling, 340–341
 - uplink shared channels, 299
 - wireless digital communication systems, 45–46
- Modulation and Coding Scheme (MCS), CQI, 322, 324
- modulation mapping
 - in downlink, 264
 - in uplink, 297–298
- modulation processing
 - downlink transport channel and, 257–258
 - layer mapping and precoding, 264–266
 - modulation mapping, 264
 - overview of, 263
 - resource mapping, 266–268
 - scrambling, 263–264
 - uplink transport channels, 297–298
- MRC (maximal ratio combining)
 - Alamouti STBC vs., 181–182
 - defined, 171–172
 - linear interference suppression and, 191–192
 - receive diversity and, 173–174
- MRT (maximal radio transmission), 191–192
- MS (mobile stations)
 - in 3G evolution of GSM, 13
 - 3GPP channel model for MIMO simulations, 78
 - channel feedback, 206–208
 - GSM base station architecture, 8–9
- MSC (Mobile Switching Center), GSM, 8–9
- MSR (Maximum Sum Rate) algorithm, 147–148, 150–152
- MTCH (Multicast Traffic Channel) logical channel, 236–237
- MTS (Mobile Telephone System), 3
- MU-MIMO (multiuser MIMO)
 - defined, 32
 - in downlink, 214, 251, 271, 275–276
 - multiuser virtual MIMO, 215
 - overview of, 209–212
 - PMI and RI feedback in, 336
 - signaling for uplink/downlink scheduling, 344
 - in uplink, 254, 301
- multi-access scheme, OFDM as, 29
- multiantenna transmission and reception. *See* MIMO (multiantenna transmission and reception)
- Multicarrier CDMA (MC-CDMA), in OFDM systems, 137–138
- multicarrier modulation
 - circular convolution and DFT, 104
 - frequency synchronization, 114–116
 - OFDM and SC-FDE computational complexity advantage and, 127–129
 - OFDM for ISI-suppression, 91, 101–104
 - OFDM in LTE, 109–110
 - often referred to as OFDM, 100
 - overview of, 99–101
 - SC-FDE for, 124–127
 - summary and conclusions, 130
 - timing synchronization, 110–114
- multicarrier modulation, OFDM basics
 - block transmission with guard levels, 103–104
 - circular convolution and DFT, 104–105
 - cyclic prefix, 104–107
 - frequency equalization, 107–108
 - steps in communication, 108–109
- multicarrier modulation, PAR (peak-to-average ratio)
 - clipping and other reduction techniques for, 121–123

LTE's approach in uplink to, 123–124
 OFDM signals having high, 116
 problem of, 116–118
 quantifying, 118–120
 Multicast/Broadcast Single Frequency Network. *See* MBSFN
 (Multicast/Broadcast Single Frequency Network)
 Multicast Channel (MCH) downlink channel, 237, 284–285
 multicast channels. *See also* MBSFN
 (Multicast/Broadcast Single Frequency Network)
 Multicast Control Channel, 236–237
 Multicast Traffic Channel, 236–237
 multicell processing (networked MIMO), 212–213
 multicell transmission, E-MBMS, 284
 multidimensional correlation, of received signal, 74–75
 multimedia, 3G systems, 10
 Multimedia Broadcast and Multicast Service. *See* MBMS (Multimedia Broadcast and Multicast Service)
 multipath, converting to flat-fading, 208
 Multipath Intensity Profile, 64
 multipath interference, 28, 60–62
 multipath modelling, 77–80
 multiple access for OFDM systems. *See also* frequency domain multiple access
 CDMA of MC-CDMA, 137–138
 FDMA, 136–137
 overview of, 134–135
 random access vs., 135–136
 TDMA, 137
 multiplexing
 Diversity-Multiplexing Tradeoff, 200–201
 spatial. *See* spatial multiplexing
 multiuser diversity
 exploiting with resource allocation, 249
 frequency diversity vs. spatial diversity vs., 159–160
 OFDMA and, 144–146
N
 N-Channel Stop-and-Wait protocol, H-ARQ, 291, 293
 N paths, multipath modelling, 77–79
 NACK_SN (Negative Acknowledgement SN) field, STATUS PDU, 366

Nakagami- m fading distribution, 70–71
 narrowband fading, mitigation of
 adaptive modulation and coding for, 86–89
 automatic repeat requests for, 86
 coding and interleaving for, 83–85
 combining techniques for, 89
 effects of unmitigated fading, 80–82
 overview of, 80
 spatial diversity for, 82–83
 narrowband interference, OFDM
 robustness against, 29
 NAS (Non-Access Stratum) protocols, 359
 Negative Acknowledgement SN (NACK_SN) field, STATUS PDU, 366
 network architecture, LTE
 design principles, 229
 GSM/GPRS, 8–9
 main components of, 231–232
 overview of, 33–35
 radio interface protocols, 232–234
 network switching, GSM for, 9
 Network Switching Subsystem (NSS), 8, 13
 networked MIMO systems, 212–213
 NLOS (non-line-of-sight) conditions
 robust ISI-suppression for, 99
 WiMAX operating in, 19
 NMT-400 (Nordic Mobile Telephone), 5
 Node-B, UMTS, 13, 32
 Node-B, UTRAN, 227
 non-3GPP systems, LTE co-existence with, 27
 Non-Access Stratum (NAS) protocols, 359
 non-GBR (Guaranteed Bit Rate) bearers, 356
 non-synchronized random access procedure, 348–350
 nonlinear equalizers, suppressing ISI, 91
 Nordic Mobile Telephone (NMT-400), 5
 NSS (Network Switching Subsystem), 8, 13
 NTACS (Total Access Communication System), 5
 NTT (Telephone and Telegraph Company), 5

O

OBO (output backoff), and PAR, 116–117, 118–120
 OC (optimum combiner), 191
 OCI (other cell interference)
 limiting wireless cellular performance, 56

- OCI (*continued*)
 mitigating with frequency reuse, 58
 mitigating with sectoring, 59–60
- OFDM (Orthogonal Frequency Division Multiplexing)
 baseband signal generation, 268
 cell-specific reference signals using, 286–287
 converting frequency-selective fading to flat-fading, 208
 DCI channel encoding and modulation, 280–281
 frequency synchronization, 110–111, 114–116
 history of, 129
 HSPA vs. WiMAX vs. LTE and, 20
 implementing FDMA, 136–137
 ISI-suppression with, 91–92
 in LTE downlink, 241–242
 LTE enabled with, 28–30, 109–110
 LTE standardization with, 228
 MIMO. *See* MIMO-OFDM
 multicarrier modulation using, 99
 multidimensional correlation of received signal in, 74–75
 multiple access for, 134–138
 PAR challenges. *See* PAR (peak-to-average ratio)
 SC-FDE vs., 124–129
 timing synchronization, 110–114
 WiMAX physical layer based on, 18–19
- OFDM (Orthogonal Frequency Division Multiplexing) basics
 block diagram of, 108–109
 block transmission with guard levels, 103
 circular convolution and DFT, 104–105
 cyclic prefix, 105–107
 frequency equalization, 107–108
 overview of, 103
- OFDM symbols
 block transmission with guard levels, 103
 and cyclic prefix, 105–107
 removing ISI within, 104–105
 timing synchronization and, 111–114
- OFDMA (Orthogonal Frequency Division Multiple Access). *See also* opportunistic scheduling for OFDMA
 adaptive modulation in, 89
 advantages and disadvantages of, 142
 how it works, 139–141
 HSPA vs. WiMAX vs. LTE, 20
 in LTE, 152–155
 LTE downlink and, 228
 LTE downlink radio resources. *See* downlink OFDMA radio resources
 multiuser diversity in, 144–146
 overview of, 138–139
 SC-FDMA vs., 142–144
 WiMAX physical layer based on, 19
- OI (Overload Indicator) messages, uplink ICI mitigation, 380
- OL (open-loop) antenna selection, LTE uplink shared channels, 301
- OL (open-loop) MIMO
 for LTE downlink shared channels, 214, 271–273
 OL spatial multiplexing supported by MU-MIMO, 251
 OL spatial multiplexing without channel feedback, 194–198
 PMI and RI feedback in, 334–335
- OL (open-loop) power control, random access preamble, 349
- open-loop transmit diversity
 with 2×1 space-frequency block coding, 175–177
 with 2×2 SFBC, 177–178
 with 4×2 in LTE, 179–180
 with 4×2 STBCs, 178–179
 defined, 174–175
- opportunistic beamforming, 159–160
- opportunistic scheduling for OFDMA
 design considerations, 155–160
 Maximum Fairness algorithm, 148
 Maximum Sum Rate algorithm, 147–148
 overview of, 146–147
 performance comparison, 150–152
 Proportional Fairness scheduling, 149–150
 Proportional Rate Constraints algorithm, 148–149
- optimum combiner (OC), 191
- optimum decoding, OL- MIMO, 194–195
- optimum eigen-beamformer, 191
- Orthogonal Frequency Division Multiple Access. *See* OFDMA (Orthogonal Frequency Division Multiple Access)

- Orthogonal Frequency Division Multiplexing. *See* OFDM (Orthogonal Frequency Division Multiplexing)
 - orthogonal (non-interfering) communication channels, 134–138
 - orthogonal space-time block code (OSTBC), 175
 - OSTBC (orthogonal space-time block code), 175
 - other cell interference. *See* OCI (other cell interference)
 - output backoff (OBO), and PAR, 116–117, 118–120
 - Overload Indicator (OI) messages, uplink ICI mitigation, 380
- P**
- P (Polling bit) field, RLC PDUs, 366
 - P-RNTI (Paging-Radio Network Temporary Identifier), 377
 - packet control unit (PCU), GSM, 8–9
 - Packet Data Convergence Protocol. *See* PDCP (Packet Data Convergence Protocol)
 - Packet Data Network Gateway (PDN GW), LTE network architecture, 232
 - Packet Data Network Gateway (PGW), EPC, 33–34
 - Packet Data Network (PDN), 356–357
 - Packet Error Rate. *See* PER (Packet Error Rate)
 - packet-switched networks, LTE support for, 229
 - padding, in trellis termination, 260
 - Paging Channel (PCH), 237
 - Paging Control Channel (PCCCH), 236, 237
 - Paging Frame (PF) function, 376–377
 - paging, mobility management in LTE, 376–377
 - Paging-Radio Network Temporary Identifier (P-RNTI), 377
 - PAPR (peak-to-average power ratio), 116, 289
 - PAR (peak-to-average ratio)
 - clipping and other reduction techniques for, 121–123
 - implementing OFDM without high, 91, 116
 - LTE's approach in uplink to, 123–124
 - problem of, 116–118
 - quantifying, 118–120
 - SC-FDMA vs. OFDMA, 144
 - Parallel Concatenated Convolutional Code (PCCC), 260–261
 - parameters, channel
 - computing effect of shadowing on, 53–54
 - key wireless, 46–48
 - LTE channels for multipath modelling, 78–79
 - OFDM in LTE, 110
 - summary of broadband fading, 63
 - path loss
 - broadband wireless channels and, 48–51
 - capacity of system increased by, 55
 - empirical/semi-empirical models estimating, 76
 - fading vs., 60
 - LTE channel models for, 76–77
 - shadowing causing fluctuations in, 51–52
 - PBCH (Physical Broadcast Channel), 239, 283
 - PCCC (Parallel Concatenated Convolutional Code), 260–261
 - PCCH (Paging Control Channel), 236, 237
 - PCFICH (Physical Control Format Indicator Channel), 239, 282
 - PCH (Paging Channel), 237
 - PCRF (Policy and Charging Rules Function)
 - EPC architecture, 33–35
 - LTE network architecture, 232
 - PCU (packet control unit), GSM, 8–9
 - PDCCH (Physical Downlink Control Channel), 350–352
 - closed-loop power control with, 154
 - DCI channel encoding/modulation with, 279–281
 - downlink control channels sent over, 276
 - Downlink Control Information sent over, 238
 - MIMO for downlink control channels over, 282–283
 - overview of, 154, 239
 - Paging function and, 377
 - random access response on, 349
 - resource allocation for uplink on, 254
 - semi-persistent scheduling for VoIP, 345–346

- PDCCH (*continued*)
 signaling for downlink scheduling, 340–341
 signaling for uplink scheduling, 341–343
- PDCP (Packet Data Convergence Protocol)
 header compression, 361
 integrity and ciphering, 361–362
 at Layer 2 of LTE protocol stack, 357–359
 overview of, 359–361
 SAPs between different layers, 234–235
- PDFs (probability density functions), 69–70, 144–146
- PDN GW (Packet Data Network Gateway), LTE network architecture, 232
- PDN (Packet Data Network), 356–357
- PDSCH (Physical Downlink Shared Channel)
 CQI estimation process, 324
 DL-SCHs carried on, 13
 MIMO for downlink control channels, 282–283
 MIMO transmission modes for, 251, 270–271
 paging information carried on, 377
 as physical channel, 239
 resource mapping of, 270
 signaling for downlink scheduling, 340–341
- PDUs (protocol data units)
 integrity protection for PDCP, 362
 MAC data formats, 367–368
 protocol architecture in W-CDMA, 359
 RLC data formats, 365–367
 RLC data transfer modes, 363–364
 RLC STATUS PDU format, 366–367
 user plane vs. control plane PDCP data, 360
- peak-to-average power ratio (PAPR), 116, 289
- peak-to-average ratio. *See* PAR (peak-to-average ratio)
- Ped (Pedestrian) A/B multipath channel models, 79
- PER (Packet Error Rate)
 effects of unmitigated fading, 82–83
 performing AMC, 87–88
 tuning AMC controller, 89
- Per Unitary basis stream User and Rate Control (PU2RC), 211
- performance
 aperiodic feedback model, 207
 implementing OFDM for graceful degradation of, 29
 LTE design requirements, 229–230
 OFDMA resource allocation schemes, 150–152
 SC-FDE vs. OFDM, 126
 SC-FDMA vs. OFDMA, 144
 wireless cellular system, 56–58
- periodic BSR, in uplink, 338
- periodic CQI reporting, 206–208, 325–328
- persistent scheduling, 344
- Personal Handyphone System (PHS), 6
- PF (Paging Frame) function, 376–377
- PF (Proportional Fairness) scheduling, 149–152
- PGW (Packet Data Network Gateway), EPC, 33–34
- PHICH (Physical Hybrid ARQ Indicator Channel)
 defined, 239
 HI for uplink transmission, 238, 321
 HI mapped onto, 282
 MIMO for downlink control channels, 282–283
- PHS (Personal Handyphone System), 6
- PHY (physical layer), LTE radio interface channel mapping at, 240–241
 downlink transport channel
 processing at. *See* downlink transport channel processing
 overview of, 234–235
 transport channels used by, 236–237
 uplink transport channel processing. *See* uplink transport channel processing
- PHY (physical layer) procedures and scheduling
 Buffer Status Reporting in uplink, 337–338
 cell search, 346–348
 CQI feedback, 322–333
 H-ARQ feedback, 319–321
 overview of, 319
 power control in uplink, 350–352
 precoder for closed-loop MIMO operations, 333–337
 random access procedures, 348–350
 scheduling and resource allocation, 339–344

- semi-persistent scheduling for VoIP, 344–346
 - summary and conclusions, 352–353
 - uplink channel sounding, 337
- PHY (physical layer), WiMAX, 18–19
- Physical Broadcast Channel (PBCH), 239, 283
- physical channels
 - actual transmission using, 239
 - defined, 234–235
 - downlink, 239
 - mapping between transport channels and, 240–241
 - signals embedded in, 240
 - uplink, 239–240
- Physical Control Format Indicator Channel (PCFICH), 239, 282
- Physical Downlink Control Channel. *See* PDCCH (Physical Downlink Control Channel)
- Physical Downlink Shared Channel. *See* PDSCH (Physical Downlink Shared Channel)
- Physical Hybrid ARQ Indicator Channel. *See* PHICH (Physical Hybrid ARQ Indicator Channel)
- physical layer, LTE radio interface. *See* PHY (physical layer), LTE radio interface
- physical layer procedures and scheduling. *See* PHY (physical layer) procedures and scheduling
- physical (PHY) layer, WiMAX, 18–19
- Physical Random Access Channel. *See* PRACH (Physical Random Access Channel)
- physical resource blocks. *See* PRBs (physical resource blocks)
- physical resource element (PRE), CQI estimation, 323
- physical signals. *See* downlink physical signals
- Physical Uplink Control Channel. *See* PUCCH (Physical Uplink Control Channel)
- Physical Uplink Shared Channel. *See* PUSCH (Physical Uplink Shared Channel)
- pilot tones, in channel estimation, 202–204
- PMCH (Physical Multicast Channel), 239
- PMI (Precoding Matrix Indicator) feedback
 - channel feedback and, 207
 - CQI reports including, 325
 - defined, 301
 - periodic CQI reporting modes and, 326–328
 - precoder for closed-loop MIMO, 334–337
 - reporting periods for CQI feedback, 330
 - UCI on PUCCH, 302–304
 - UCI on PUSCH with UL-SCH data, 305
 - wideband CQI reporting, 329–330
- PO (Paging Occasion), 376–377
- Policy and Charging Rules Function (PCRF)
 - EPC architecture, 33–35
 - LTE network architecture, 232
- Polling bit (P) field, RLC PDUs, 366
- postcoding, 198–200
- power control
 - cyclic prefix creating loss of, 106–107
 - LTE and, 154–155
 - random access preamble transmission and, 349
 - spatial diversity reducing transmit, 171
 - tuning AMC controller, 89
 - in uplink, 350–352
 - uplink ICI mitigation, 379
- power delay profile, 64
- PRACH (Physical Random Access Channel)
 - defined, 239
 - generating random access, 314
 - random access preamble transmission on, 349
- PRBs (physical resource blocks)
 - for CQI estimation, 323
 - for OFDMA, 246–250
 - for SC-FDMA, 252–254
 - for subband CQI reports, 322
- PRC (Proportional Rate Constraints) algorithm, 148–149, 151–152
- PRE (physical resource element), CQI estimation, 323
- preambles, channel estimation, 202–204
- precoder for closed-loop MIMO
 - estimation for multicarrier systems, 333–334
 - overview of, 333
 - PMI and RI feedback, 334–337

precoding
 codebook, 206
 layer mapping and, 264–266
 quantized feedback for linear, 206–207
 rank-1, 274–275

Precoding Matrix Indicator. *See* PMI
 (Precoding Matrix Indicator)
 feedback

preparation phase
 S1 mobility, 371–372
 X2 mobility, 373–374

primary synchronization signals, 289–290

probability density functions (PDFs),
 69–70, 144–146

profiles, ROHC, 361–362

Proportional Fairness (PF) scheduling,
 149–152

Proportional Rate Constraints (PRC)
 algorithm, 148–149, 151–152

protocol architecture in LTE
 Layer 2, 357–358
 LTE radio interface, 232–234
 MAC/RLC. *See* MAC/RLC overview
 overview of, 356–358
 PDCP, 357–362
 RRC, 369–371
 in W-CDMA, 359

protocol data units. *See* PDUs (protocol
 data units)

PSTN (Public Switched Telephone
 Network), 9

PU2RC (Per Unitary basis stream User
 and Rate Control), 211

Public Switched Telephone Network
 (PSTN), 9

PUCCH (Physical Uplink Control
 Channel)
 CQI feedback modes on, 325–326
 defined, 239
 H-ARQ transmissions in downlink
 and, 320
 modulation of, 306–308
 periodic CQI reporting, 326–330
 resource mapping of, 308–309
 resource mapping of demodulation
 reference signals, 310–312
 UCI on, 239, 302–304

punctured codes, LTE, 42, 84–85

PUSCH (Physical Uplink Shared Channel)
 aperiodic CQI reporting in, 331–332
 CQI feedback modes on, 325–326
 defined, 239
 frequency hopping on, 299–300

power control in uplink for
 transmissions of, 350–352
 resource mapping of, 308–309
 resource mapping of demodulation
 reference signals, 310–311
 signaling for uplink scheduling,
 341–343
 UCI on, 239
 UCI on, with UL-SCH data,
 304–306
 UCI on, without UL-SCH data, 306
 for Uplink Control Information, 302
 uplink shared channels mapped
 to, 299

Q

Q function, 54, 58

QCI (QoS Class Identifier), 338, 356–357

QoS (Quality of Service), 11, 355–359

QPP (Quadrature Permutation
 Polynomial) interleaver, 260–262

QPSK modulation
 AMC, 87–88
 for broadcast channels, 283
 CQI estimation process, 324–325
 downlink shared channels, 269
 LTE schemes for, 264
 multiuser diversity of OFDMA and,
 145
 uplink shared channels, 299

Qualcomm, 9

R

R field, MAC subheader, 367

RA-RNTI (Random Access Radio Network
 Temporary Identifier), 349

RABs (Radio Access Bearers), 372–374

RACH (Random Access Channel)
 overview of, 237
 random access preamble transmission,
 349
 Si mobility, 372

radio
 development of shortwave, 3
 inventors of, 3
 link quality in LTE, 374–375
 UTRAN/E-UTRAN architectures,
 227–228

Radio Access Bearers (RABs), 372–374

radio bearers
 mapped to RLC and PDCP, 358
 overview of, 355–356
 PDCP used for, 357

- Radio Link Control. *See* RLC (Radio Link Control) layer
- radio network temporary identifier (RNTI), 279
- Radio Resource Control. *See* RRC (Radio Resource Control) layer
- radio resources
 - downlink OFDMA. *See* downlink OFDMA radio resources
 - LTE management requirements, 230
 - uplink SC-FDMA. *See* uplink SC-FDMA radio resources
- RAKE receivers, 9, 90
- RAN (Radio Access Network)
 - GSM network architecture, 8
 - LTE radio interface and, 227–228
 - procedures for mobility, 374–376
- RAN (Radio Access Network) group, LTE project, 23–24
- random access
 - multiple access vs., 135–136
 - preamble, 313–314, 349
 - procedures, 348–350
 - response, 368
 - response grant, 349–350
 - in uplink, 313–314
- Random Access Channel. *See* RACH (Random Access Channel)
- Random Access Radio Network Temporary Identifier (RA-RNTI), 349
- random orthogonal beamforming, 211
- randomization, ICI, 377, 379
- range, bandwidth vs., 49
- Rank Indication. *See* RI (Rank Indication) feedback
- rate matching, 262–263, 269
- Rayleigh fading model, 68–69, 71–75
- Re-segmentation Flag (RF), RLC PDUs, 366
- receive antenna arrays, mitigating narrowband fading, 83
- receive diversity
 - maximal ratio combining in, 173–174
 - overview of, 171
 - selection combining in, 171–173
 - transmit diversity vs., 180–181
- received signal
 - selection diversity and, 83
 - statistical correlation of, 71–75
- receivers, multicarrier
 - concept of, 101–102
 - demodulating OFDM signal, 110–111
 - OFDM/SC-FDE and, 125–127
- Reference Signal Code Power (RSCP), radio link quality, 374–375
- Reference Signal Received Power (RSRP), radio link quality, 374–375
- reference signals
 - in downlink, 240, 285–288
 - in uplink, 240, 309–313
- regular BSR, in uplink, 338
- Relative Narrowband Transmit Power (RNTP) indicator, ICI coordination/avoidance, 378
- reporting periods, CQI feedback, 330
- resource allocation
 - downlink OFDMA, 248–251
 - LTE channel dependent multi-user, 30–31
 - non-synchronized random access procedure, 348–349
 - OFDMA design, 155–160
 - OFDMA opportunistic scheduling and, 146–147
 - role of, 248
 - signaling for downlink scheduling, 340–341
 - WiMAX per user, 19
- resource allocation, scheduling and MU-MIMO signaling, 344
- overview of, 339
- signaling for downlink scheduling, 340–341
- signaling for uplink scheduling, 341–343
- resource block groups (RBGs), 249–250
- resource blocks
 - LTE time-frequency grid and, 153–154
 - modulation processing in uplink, 298
 - for OFDMA, 246–248
 - for OFDMA, resource allocation, 248–249
 - for SC-FDMA, 252–254
 - specifying allocation of downlink with PDCCH, 154
 - UE-specific reference signals transmitted on, 287
- resource elements
 - cell-specific reference signals, 287
 - in downlink resource grids, 246, 248
 - in uplink resource grids, 252–254
- resource grid structure
 - downlink, 246–248
 - uplink, 252–253

- resource mapping
 - of cell-specific reference signals, 287
 - of demodulation reference signals, 310–312
 - of MBSFN reference signals, 287–288
 - in modulation processing, 266–268
 - of PDSCH physical channel, 270
 - of sounding reference signals, 311–313
 - of UE-specific reference signals, 288
 - of uplink transport channels, 308–309
 - Rev. A, EV-DO enhancement, 13
 - RF (Re-segmentation Flag), RLC PDUs, 366
 - RF signal, 46
 - RGBs (resource block groups), 249–250
 - RI (Rank Indication) feedback
 - CQI reports including, 325
 - defined, 301
 - precoder for closed-loop MIMO, 334–337
 - reporting periods for CQI feedback, 330
 - UCI on PUSCH with UL-SCH data, 305
 - Ricean distribution, statistical fading model, 69–70
 - RLC (Radio Link Control) layer. *See also* MAC/RLC overview
 - handling H-ARQ error events, 292
 - HSPA+ enhancements to, 17–18
 - Layer 2 of LTE protocol stack, 357–358
 - logical channel services to, 235–236
 - LTE, 234
 - PDCP used for, 357
 - SAPs between different layers, 234–235
 - RNC (Radio Network Controller)
 - in 3G evolution of GSM, 9, 13
 - E-UTRAN network architecture vs., 227
 - evolution of 3GPP network architecture, 32
 - protocol architecture in W-CDMA, 359
 - RNTI (radio network temporary identifier), 279
 - RNTP (Relative Narrowband Transmit Power) indicator, ICI coordination/avoidance, 378
 - ROHC (RObust Header Compression) protocol
 - LTE header compression with, 361–362
 - PDCP based on, 233
 - PDCP for user plane and, 357
 - round-robin scheduling. *See also* TDMA (Time Division Multiple Access) systems
 - comparing OFDMA resource allocation, 150–152
 - defined, 137
 - RRC (Radio Resource Control) layer
 - configuring H-ARQ transmissions at, 292
 - EPS bearer service architecture, 355–356
 - functions of, 369–371
 - overview of, 232–233
 - paging function of, 376–377
 - semi-persistent scheduling for VoIP at, 345–346
 - states, 369–370
 - RRC_CONNECTED state, 369–370, 374–376
 - RRC_IDLE state, 369–370, 374–376
 - RSCP (Reference Signal Code Power), radio link quality, 374–375
 - RSRP (Reference Signal Received Power), radio link quality, 374–375
- ## S
- S-GW (Serving Gateway)
 - EPC architecture, 33–34, 355–356
 - LTE network architecture, 232
 - X2 mobility, 373–374
 - S1 bearer, EPS, 355–356
 - S1 interface, LTE network architecture, 232
 - S1 mobility, management in LTE, 371–373
 - S5/S8 bearer, EPS, 355–356
 - SAE-GW (System Architecture Evolution Gateway), LTE, 32–33
 - SAE (Systems Architecture Evolution) project, 24
 - safety, early mobile communications, 3
 - SAPs (Service Access Points)
 - radio interface protocols and, 234–235
 - for services to MAC and RLC, 358
 - SC-FDE (Single Carrier Frequency Domain Equalization)
 - computational complexity advantage of, 127–128

- for core modulation in uplink, 30, 123–124
 - design considerations, 126–127
 - ISI mitigation, 91
 - OFDM performance vs., 126
 - system description, 124–125
- SC-FDMA (Single Carrier Frequency Division Multiple Access)
 - advantages and disadvantages of, 143–144
 - how it works, 142–143
 - for ISI-suppression, 91
 - in LTE, 152–155
 - LTE using, 20, 30
 - modulation processing in uplink, 297–298
 - multicarrier modulation
 - underlying, 99
 - overview of, 142
 - standardizing LTE in uplink with, 228
 - uplink LTE radio resources. *See* uplink SC-FDMA radio resources
- SC (selection combining), in receive diversity, 171–173
- scalability, 19, 55
- SCH (Supplemental Code Channel), CDMA, 10
- scheduling
 - physical layer procedures and. *See* physical layer procedures and scheduling
 - resource allocation and. *See* resource allocation, scheduling and VoIP semi-persistent, 344–346
- scheduling algorithms, OFDMA
 - approaches, 146–147
 - Maximum Fairness algorithm, 148
 - Maximum Sum Rate algorithm, 147–148
 - overview of, 146–147
 - performance comparison, 150–152
 - Proportional Fairness scheduling, 149–150
 - Proportional Rate Constraints algorithm, 148–149
- Scheduling Request. *See* SR (Scheduling Request)
- scrambling
 - mitigating ICI with, 377
 - modulation in downlink, 263–264
 - modulation in uplink, 297–298
 - modulation of PUCCH, 307–308
 - uplink shared channels, 299
- SDMA (Space Division Multiple Access), MU-MIMO and, 210–212
- SDU (service data unit), W-CDMA, 359
- second generation digital systems. *See* 2G (second generation) digital cellular systems
- secondary synchronization signals, 289–290
- sectoring, improving SIR with, 58–60
- security
 - 2G systems improving, 6
 - integrity and ciphering in PDCP, 361–362
- Segment Offset (SO) field, RLC PDUs, 366
- selection combining (SC), in receive diversity, 171–173
- selection diversity, 83
- selective retransmission, 374
- selectivity, dispersion-selectivity duality, 74–75
- self-interference, in OFDMA and SC-FDMA, 154–155
- semi-empirical channel models, LTE, 79–80
- semi-persistent scheduling for VoIP, 344–346
- semi-static ICI coordination/avoidance, 378
- Sequence Number field. *See* SN (Sequence Number) field
- Service Access Points (SAPs)
 - radio interface protocols and, 234–235
 - for services to MAC and RLC, 358
- service data unit (SDU), W-CDMA, 359
- Serving Gateway. *See* S-GW (Serving Gateway)
- Serving GPRS Support Node. *See* SGSN (Serving GPRS Support Node)
- Serving Radio Network Subsystem (SRNS) relocation, 371
- SFBC (space-frequency block code)
 - 2×1 transmit diversity, 175–177
 - 2×2 transmit diversity, 177–178
 - LTE transmit diversity based on, 31, 179–180
 - multiuser diversity vs. frequency diversity vs. spatial diversity in, 159–160
 - in OL MIMO for transmit diversity, 272
 - STBCs vs., 175

- SFBC (*continued*)
 - SU-MIMO supporting, 251
 - transmit diversity vs. receive diversity and, 181
- SFN (Single-Frequency Network) mode
 - HSPA+ enhancements to, 18
 - implementing OFDM for broadcast services, 30
 - LTE support for E-MBMS through, 284
- SFR (Soft Frequency Reuse), mitigating ICI with, 157–158
- SGSN (Serving GPRS Support Node)
 - defined, 232
 - evolution of 3GPP, 32
 - GSM network architecture, 8–9
 - MME vs., 231
- shadowing
 - affect on wireless system, 52–55
 - empirical/semi-empirical models
 - estimating, 76
 - fading vs., 60
 - overview of, 51–52
- Shannon capacity formula, 22, 147–148, 170–171
- shared channels
 - downlink. *See* DL-SCHs (Downlink Shared Channels)
 - transmission, 237
 - uplink, 298–301
- Short Messaging Service (SMS) messages, 2G, 6–7
- shortwave radios, 3
- SIBs (System Information Blocks)
 - broadcast channels, 283
 - cell search process, 348
 - functions of RRC protocol, 371
- signal-to-interference plus noise ratio. *See* SINR (signal-to-interference plus noise ratio)
- signal-to-interference ratio. *See* SIR (signal-to-interference ratio)
- signal-to-noise-plus-distortion ratio (SNDR), 122–123
- signal-to-noise ratio. *See* SNR (signal-to-noise ratio)
- Signaling Radio Bearers. *See* SRBs (Signaling Radio Bearers)
- signals. *See also* downlink physical signals
 - downlink scheduling, 340–341
 - for downlink/uplink physical layers, 240
 - generating OFDM baseband, 268
 - HSPA vs. WiMAX vs. LTE, 20–21
 - for MU-MIMO in uplink, 344
 - time-continuous random access, 314
 - uplink scheduling, 341–343
- single antenna port
 - downlink shared channels, 270–271
 - layer mapping and precoding, 264–266
- Single Carrier Frequency Division Multiple Access. *See* SC-FDMA (Single Carrier Frequency Division Multiple Access)
- Single Carrier Frequency Domain Equalization. *See* SC-FDE (Single Carrier Frequency Domain Equalization)
- single-cell transmission, E-MBMS, 284
- Single-Frequency Network mode. *See* SFN (Single-Frequency Network) mode
- Single-Input-Single-Output. *See* SISO (Single-Input-Single-Output) mode
- single-user OFDM. *See* OFDM (Orthogonal Frequency Division Multiplexing)
- singular value decomposition (SVD), precoding and postcoding, 198
- SINR (signal-to-interference plus noise ratio)
 - analysis of cellular systems, 57
 - calculating, 51
 - comparing OFDMA resource allocation schemes, 151–152
 - CQI estimation process and, 323–325
 - CQI feedback mechanism for LTE, 322
 - effect of shadowing on, 52
 - fractional frequency reuse and, 157
 - maximizing total throughput with MSR, 147–148
 - OFDMA architecture, 138–142
 - performing AMC, 87–88
 - power control in uplink and, 350–352
 - tuning AMC controller, 88–89
- SIR (signal-to-interference ratio)
 - analysis of cellular systems, 56–58
 - calculating using empirical path loss formula, 50–51
 - improving with sectoring, 58–60
- SISO (Single-Input-Single-Output) mode
 - channel feedback in, 206
 - for downlink shared channels, 271
 - multiuser diversity vs. frequency diversity vs. spatial diversity in, 159–160
- slotted ALOHA, in random access, 135

- smart mobile devices, LTE
 - development, 25
- SMS (Short Messaging Service) messages, 2G, 6–7
- SN (Sequence Number) field
 - ACK_SN field, STATUS PDU, 366
 - NACK_SN field, STATUS PDU, 366
 - PDCCP, 360–362
 - RLC PDUs, 365
- SNDR (signal-to-noise-plus-distortion ratio), 122–123
- SNR (signal-to-noise ratio)
 - array gain and, 169
 - diversity gain and, 169–170
 - effects of shadowing, 53
 - effects of unmitigated fading, 80–81
 - increasing data rates with spatial diversity, 170–171
 - linear diversity precoding and, 186
 - loss induced by frequency synchronization, 115–116
 - loss induced by timing synchronization, 113
 - maximal ratio combining maximizing, 173–174
 - nonlinear equalizers for
 - ISI-suppression and, 91
 - OFDM vs. SC-FDE performance, 126
 - selection combining and, 172
 - selection diversity increasing, 83
 - in single-user MIMO system, 194
 - spatial multiplexing and, 194
 - transmit diversity vs. receive diversity and, 180–181
- SO (Segment Offset) field, RLC PDUs, 366
- SOend (SO end) field, STATUS PDU, 366
- Soft Frequency Reuse (SFR), mitigating
 - ICI with, 157–158
- soft-handoff, IS-95 CDMA enabling, 9
- soft limiting (clipping)
 - OFDM/SC-FDE design, 127
 - reducing PAR with, 121–123
- SOstart (SO start) field, STATUS PDU, 366
- sounding reference signals, 309, 311–313
- Sounding Reference Symbol (SRS), 337
- Space Division Multiple Access (SDMA), MU-MIMO and, 210–212
- space-frequency block code. *See* SFBC (space-frequency block code)
- space-time block codes (STBCs), 175, 178–179
- space-time trellis codes, 177
- spatial diversity
 - array gain and, 168–169
 - diversity gain and, 169–170
 - frequency vs. multiuser vs., 159–160
 - increasing coverage area with, 171
 - increasing data rate with, 170–171
 - interference suppression vs. spatial multiplexing vs., 200–202
 - mitigating narrowband fading, 83–84
 - overview of, 83–84, 168
 - receive diversity as, 171–174
 - reducing transmit power with, 171
 - transmit diversity. *See* transmit diversity
- spatial multiplexing
 - diversity vs. interference suppression vs., 200–202
 - introduction to, 193–194
 - layer mapping and precoding, 266
 - overview of, 192–193
 - using transmit selection with, 183
- spatial multiplexing, CL MIMO
 - downlink shared channels, 274–275
 - LTE downlink, 214
 - overview of, 198–200
 - PMI and RI feedback in, 336
- spatial multiplexing, OL MIMO
 - downlink shared channels, 271–272
 - overview of, 214
 - PMI and RI feedback in, 336
 - without channel feedback, 194–198
- spectral dispersion, OFDM/SC-FDE
 - design, 127
- spectral regrowth
 - reducing PAR with clipping causing, 121–123
 - SC-FDMA disadvantages, 144
- spectrum
 - 3G systems, 10–11
 - 4G systems, 40
 - AMPS, 6
 - concept in 1970s cellular mobiles, 3
 - early mobile telephone systems and, 3
 - LTE-Advanced requirements, 40
 - LTE design principles, 26–27, 229–230
 - OFDM and SC-FDE design, 127
 - options for LTE, 35–39
- spread spectrum
 - ISI suppression using, 99
 - RAKE receivers and, 90
- SR (Scheduling Request)
 - encoding H-ARQ-ACK and, 303–304

- SR (*continued*)
 PUCCH formats, 307
 and UCI, 301
- SRBs (Signaling Radio Bearers)
 defined in LTE, 370
 EPS bearer service architecture,
 355–356
 functions of RRC protocol, 371
- SRNS (Serving Radio Network Subsystem)
 relocation, 371
- SRS (Sounding Reference Symbol), 337
- stacked STBCs, 178–179
- standalone deployment scenario, LTE, 230
- states, RRC, 369–370
- static ICI coordination/avoidance, 378
- static TDMA. *See* round-robin scheduling
- statistical channel models
 correlation of received signal, 71–75
 defined, 67
 line-of-sight channels and, 69–70
 Nakagami-m fading distribution,
 70–71
 overview of, 68
 pedagogy for developing, 67
 Rayleigh fading model, 68–69
- STATUS PDU format, RLC, 366–367
- STBCs (space-time block codes), 175,
 178–179
- Strict FFR (Fractional Frequency Reuse),
 mitigating ICI, 157–158
- SU-MIMO (single-user MIMO)
 downlink transmission support in
 LTE, 251
 multiuser MIMO vs., 209–212
- sub-block interleaving, rate matching
 performing, 262–263
- subband CQI reporting
 CQI estimation process, 323–324
 enabling frequency selective
 scheduling, 340–341
 overview of, 322
 periodic/aperiodic CQI modes
 supporting, 325–326
 periodic CQI reports supporting,
 326–330
 reporting periods, 330
- subband feedback, CQI estimation, 324
- subcarrier bandwidth, 100–102
- subscribers, fixed line vs. mobile, 1
- supplemental channels, CDMA2000-1X
 and, 12
- Supplemental Code Channel (SCH),
 CDMA, 10
- SVD (singular value decomposition),
 precoding and postcoding, 198
- symbol time
 in ISI suppression, 100–101
 multicarrier modulation and, 91
 severity of ISI and, 63
- synchronization signals, 289–290
 demodulating OFDM signal with,
 110–112
 in downlink physical channels, 240,
 289–290
 OFDM and SC-FDE design
 considerations, 127
- synchronized random access, 348
- System Architecture Evolution Gateway
 (SAE-GW), LTE, 32–33
- System Information Blocks. *See* SIBs
 (System Information Blocks)
- Systems Architecture Evolution (SAE)
 project, 24
- ## T
- tail-biting convolutional coding,
 259–260
- tapped-delay line. *See* TDL (tapped-delay
 line) channel model
- TBS (Transport Block Size) index,
 downlink scheduling, 340–341
- TD-SCDMA (Time Division, Synchronous
 CDMA), 11
- TDD (Time Division Duplexing) mode
 aperiodic CQI reporting in, 331–332
 channel feedback in, 206
 designated frequency bands for LTE,
 35, 37
 frame structure type 2 for,
 244–246
 H-ARQ Indicator for uplink
 transmission, 315–317, 321
 H-ARQ transmissions in downlink,
 291–292, 320–321
 IMT-TC standard for, 11
 LTE designed to operate in, 230
 LTE support for, 27
 resource mapping of sounding
 reference signals, 311–313
 UCI on PUSCH with UL-SCH data,
 304–305
 W-CDMA specified for, 13–14
 WiMAX support for, 19
- TDL (tapped-delay line) channel model
 broadband fading channels and,
 61–62

- broadband wireless channels and, 47–48
 - free-space path loss (Friis) formula and, 49
- TDM, WiMAX resource allocation, 19
- TDMA (Time Division Multiple Access) systems
 - as 2G system, 6–7
 - advantages of OFDMA vs., 142
 - CDMA technology vs., 9–10
 - classifying CSMA systems as type of, 136
 - digital evolution of GSM and, 8
 - HSPA vs. WiMAX vs. LTE, 20
 - implementing in OFDM systems, 137
 - overview of, 134–135
- Telephone and Telegraph (NTT) Company, 5
- third generation broadband wireless. *See* 3G (third generation) broadband wireless systems
- Third Generation Partnership Project. *See* 3GPP (Third Generation Partnership Project)
- Third Generation Partnership Project 2 (3GPP2), 11–12
- throughputs
 - comparing OFDMA resource allocation schemes, 150–152
 - LTE design for high, 26–27
- TIA (Telecommunications Industry Association), 9
- time
 - correlation of received signal, 71–72
 - dispersion-selectivity in, 74
- time-continuous random access signal, 314
- time dimension, broadband fading channels, 62–63
- time diversity, coding and interleaving as, 83
- Time Division - Synchronous CDMA (TD-SCDMA), 11
- Time Division Duplexing. *See* TDD (Time Division Duplexing) mode
- Time Division Multiple Access. *See* TDMA (Time Division Multiple Access) systems
- time domain
 - cell-specific reference signals, 286–287
 - frame structure in, 242–243
 - ISI suppression using equalization of, 99
 - MBSFN reference signals, 287
 - resource allocation exploiting channel variations in, 249
- time-domain channel estimation, MIMO-OFDM, 203–205
- time-frequency grid, 153
- time-frequency uncertainty principle, 65
- TimeToTrigger parameter, RAN mobility management, 375
- timing synchronization
 - demodulating OFDM signal with, 110–112
 - margin, 113
 - overview of, 111–114
- TM (Transparent Mode), RLC protocol, 234, 363
- TMD (TM Data) PDUs, 363, 365
- Total Access Communication System (ETACS), 5
- Total Access Communication System (NTACS), 5
- TPMI (Transmit Precoding Matrix Indication), 336–337
- transmission modes, downlink, 270–271, 326–327
- transmission rate, mitigating multipath fading, 90
- Transmission Time Interval. *See* TTI (Transmission Time Interval)
- transmit beamforming, 159–160
- transmit diversity
 - closed-loop, 181–186
 - defined, 31
 - downlink shared channels, 270
 - MIMO supporting in LTE downlink, 213–214
 - mitigating narrowband fading with, 83–84
 - overview of, 174–175
 - physical channels supporting, 251
 - receive diversity vs., 180–181
 - SU-MIMO supporting, 251
- transmit diversity, OL MIMO
 - downlink control channels, 282–283
 - downlink shared channels, 271–272
 - layer mapping and precoding, 266
 - overview of, 175–180
- transmit power, spatial diversity reducing, 171
- Transmit Precoding Matrix Indication (TPMI), 336–337
- transmit selection diversity (TSD), 182–185

- transmitters, multicarrier
 - OFDM and SC-FDE design, 126–127
 - OFDM vs. SC-FDE, 125
 - overview of, 101–102
 - Transparent Mode (TM), RLC protocol, 234, 363
 - Transport Block Size (TBS) index,
 - downlink scheduling, 340–341
 - transport blocks, 238
 - transport channels. *See also* downlink transport channel processing; uplink transport channel processing
 - defined, 234–235
 - defining control information in MAC layer, 238
 - downlink and uplink, 237
 - how to transmit, 236–237
 - mapping between logical channels and, 240–241
 - mapping between physical channels and, 240–241
 - transport blocks, 238
 - trellis termination, tail-biting
 - convolutional coding and, 260
 - TSD (transmit selection diversity), 182–185
 - TTI (Transmission Time Interval)
 - defined, 238
 - downlink OFDMA radio resources and, 242
 - frame structure in time domain, 242–243
 - frame structure type 1 for FDD and, 243
 - for PBCH, 283
 - semi-persistent scheduling for VoIP, 346
 - signaling for downlink scheduling, 340–341
 - turbo coding
 - convolution, 259–262
 - defined by LTE for use in uplinks and downlinks, 84
 - interleaving used with, 85
 - performing AMC, 87–88
 - two-dimensional autocorrelation,
 - broadband fading channels, 61–63
 - type 0 resource allocations, 249–250
 - type 1 frame structure, 243–244, 252
 - type 1 resource allocations, 250
 - type 2 frame structure, 244–246, 252
 - Type 2 hopping, 300
 - type 2 resource allocations, 250–251
 - Typical Urban delay profile, 79
- ## U
- UCI (Uplink Control Information)
 - channel coding for, 296–297, 302–306
 - defined in MAC layer, 239
 - modulation of PUCCH, 306–308
 - overview of, 301–302
 - resource mapping, 308–309
 - UE-selected subband CQI report
 - aperiodic CQI reports supporting, 332–333
 - defined, 325
 - periodic CQI reports supporting, 327–329
 - UE-specific downlink reference signals, 287–288
 - UE (User Equipment)
 - defined, 13
 - location of radio interface, 227
 - LTE network architecture, 231
 - MAC entities in, 364
 - resource allocation and, 248–249
 - resource allocation in uplink with, 254
 - UHF spectrum, LTE deployments, 35–36, 38–39
 - UL-SCHs (Uplink Shared Channels)
 - channel coding processing, 296–297
 - defined, 237
 - overview of, 298–301
 - UCI on PUSCH with, 304–306
 - ULA (uniform linear array), DOA-based beamsteering, 187–189
 - UM (Unacknowledged Mode), RLC, 234, 363
 - UMB (Ultra Mobile Broadband), 16
 - UMD (UM Data) PDUs, 363, 365
 - UMTS Release 99, 13
 - UMTS-TDD, 11
 - UMTS Terrestrial Radio Access Network. *See* UTRAN (UMTS Terrestrial Radio Access Network)
 - UMTS Terrestrial Radio Access (UTRA), 227
 - UMTS (Universal Mobile Telephone Service)
 - LTE network architecture vs., 231
 - packet-switched network support in, 229

- using 3GPP releases, 22
 - W-CDMA standard for, 11, 13–14
 - uncorrelated antennas, MIMO, 208–209
 - uniform linear array (ULA), DOA-based beamsteering, 187–189
 - uplink
 - allocation notification/feedback in, 154
 - Buffer Status Reporting in, 337–338
 - CDMA2000-1X improvements to, 12
 - channel sounding in, 337
 - designated FDD frequency bands for LTE in, 35–36
 - designing CQI feedback for, 322
 - HSPA+ enhancements to, 16–17
 - HSPA vs. WiMAX vs. LTE in, 20–21
 - ICI coordination in LTE, 379–380
 - LTE design principles, 26, 229
 - LTE using OFDMA in, 139–141
 - LTE using SC-FDE in, 30, 123–129
 - LTE using SC-FDMA in, 142–143, 152–155
 - MIMO in LTE, 214–215
 - power control in, 154–155, 350–352
 - SDMA vs. SU-MIMO in, 211
 - signaling for scheduling in, 341–343
 - Uplink Control Information. *See* UCI (Uplink Control Information)
 - uplink feedback, LTE, 154
 - uplink physical channels, 239–240
 - Uplink Pilot TimeSlot (UpPTS) field
 - downlink OFDMA radio resources, 245
 - uplink SC-FDMA radio resources, 252
 - uplink reference signals
 - overview of, 309–310
 - resource mapping of demodulation reference signals, 310–311
 - resource mapping of sounding reference signals, 311–313
 - sequence of, 310
 - uplink SC-FDMA radio resources
 - frame structure, 252
 - overview of, 251–252
 - physical resource blocks for SC-FDMA, 252–254
 - resource allocation, 254–255
 - Uplink Shared Channels. *See* UL-SCHs (Uplink Shared Channels)
 - uplink transport channel processing
 - channel coding processing, 296–297
 - H-ARQ in uplink, 315–317
 - modulation processing, 297–298
 - overview of, 295–296
 - random access channels, 313–314
 - summary and conclusions, 317
 - uplink control information, 301–309
 - uplink reference signals, 309–313
 - uplink shared channels, 298–301
 - UpPTS (Uplink Pilot TimeSlot) field
 - downlink OFDMA radio resources, 245
 - uplink SC-FDMA radio resources, 252
 - User Equipment. *See* UE (User Equipment)
 - user-plane latency, LTE design, 229
 - user plane protocol stack
 - ciphering in PDCP for, 362
 - overview of, 356–357
 - PDCP services and functions for, 359–361
 - UTRA (UMTS Terrestrial Radio Access), 227
 - UTRAN (UMTS Terrestrial Radio Access Network). *See also* E-UTRAN (Enhanced UTRAN)
 - defined, 13
 - LTE deployment scenario requirements and, 230
 - LTE designed to operate in, 230
 - UWC-136, 11
- ## V
- V-BLAST (vertical BLAST), open-loop MIMO, 196–198
 - Veh (Vehicular) A/B multipath channel models, 79
 - video, mobile, 24–25
 - virtual MIMO transmission. *See* MU-MIMO (multiuser MIMO)
 - virtual resource blocks (VRBs), resource allocation and, 250–251
 - Viterbi decoder, 84
 - VLR (Visitor Location Register), GSM, 8–9
 - voice service
 - 2G systems improving, 6
 - 3G systems and, 10
 - IS-95 CDMA systems improving, 9–10
 - using VoIP in LTE to provide, 344–346
 - VoIP (voice over IP)
 - header compression in, 361
 - HSPA vs. WiMAX vs. LTE, 22

VoIP (*continued*)

- semi-persistent scheduling for, 344–346
- using Unacknowledged Mode of RLC, 363
- utilizing periodic feedback mode, 207
- wideband CQI reporting for, 322
- WiMAX network using, 18

VRBs (virtual resource blocks), resource allocation and, 249–251

W

W-CDMA (wideband CDMA)

- in 3G systems, 11–12
- HSPDA vs., 14–15
- overview of, 13–14
- protocol architecture in LTE, 359

Wide Sense Stationary Uncorrelated

Scattering (WSSUS), 63

Wide Sense Stationary (WSS), 63

wideband CQI reporting

- aperiodic CQI modes supporting, 325–326, 331
- CQI estimation process, 323–324
- overview of, 322
- periodic CQI modes supporting, 325–328
- reporting periods, 330
- understanding, 329–330

WiMAX

- as 4G system, 16
- evolution beyond 3G, 15
- HSPA+ vs. LTE vs., 20–22
- IMT-2000 IP-OFDMA standard as, 11
- LTE co-existence with, 27
- Mobile, 18–20
- multicarrier modulation underlying, 99

WiMAX (Worldwide Interoperability for Microwave Access) Forum, 18

wireless fundamentals

- broadband wireless channel, fading, 60–67
- broadband wireless channel, overview, 46–48

broadband wireless channel, path loss, 48–51

broadband wireless channel, shadowing, 51–55

cellular systems, 54–60

communication system building blocks, 45–46

mitigation of broadband fading, 90–92

mitigation of narrowband fading, 80–89

modelling broadband fading channels, 67–80

overview of, 45

summary, 92–93

wireless telegraphy, development of, 3

WLANs (wireless LANs)

multicarrier modulation

underlying, 99

using CSMA for random access, 135–136

WMAN (wireless metropolitan area network), standard for, 18

WSS (Wide Sense Stationary), 63

WSSUS (Wide Sense Stationary Uncorrelated Scattering), 63

X

X2 interface

- LTE network architecture, 232
- mobility, 373–374

Z

Zadoff-Chu sequence

- generating random access preambles, 313
- modulation of PUCCH, 307
- primary synchronization signals, 289
- random access preamble transmission, 349
- reference signal sequence, 310

ZF (zero-forcing) detector, OL-MIMO, 195–196